

300 meters on installed MMF

Part I: Architectures

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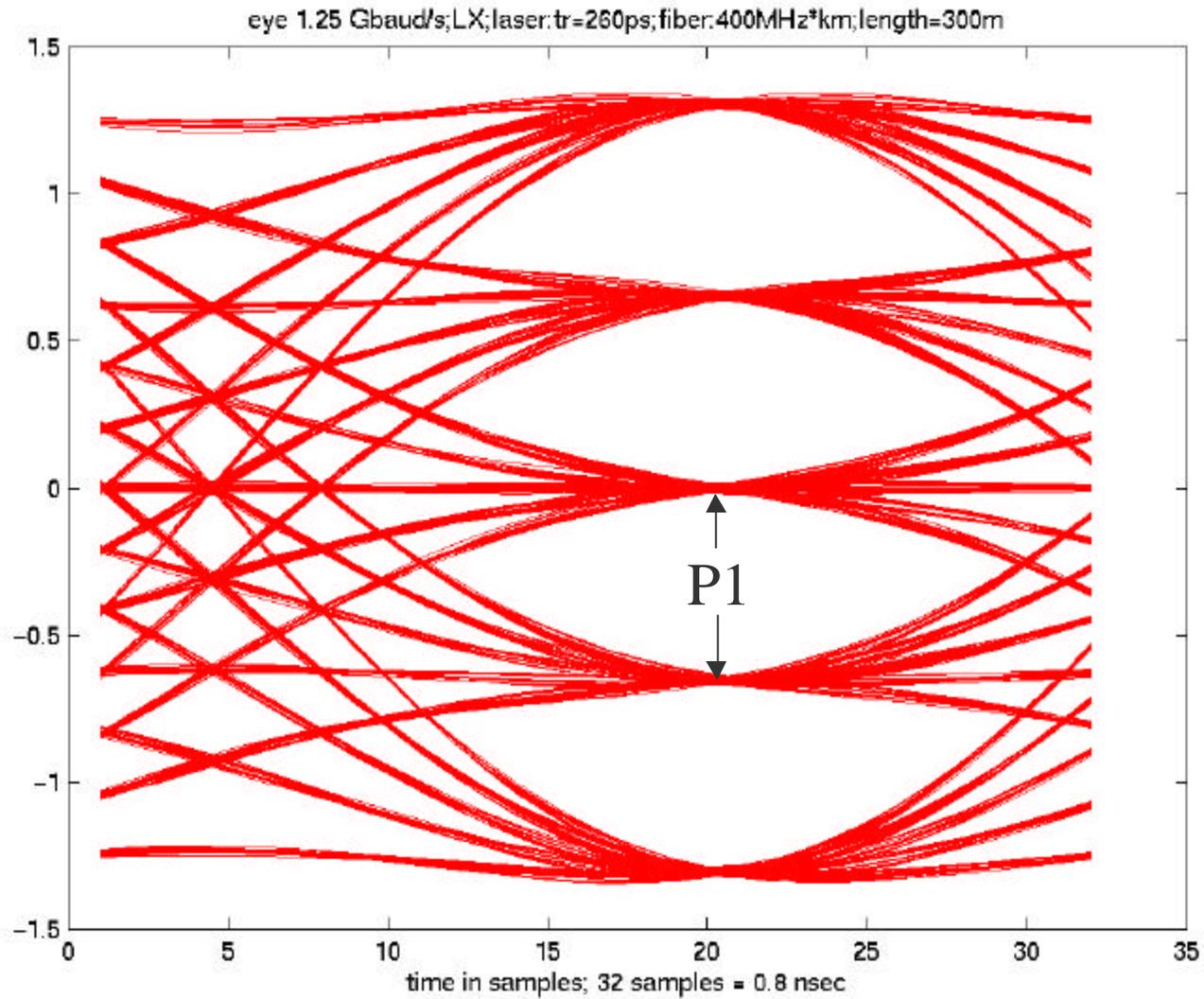
Objectives

- 300 meters provides an optimum coverage of installed multimode fiber
- compare proposed architectures that could support up to 300 meter link lengths on 50 μm and 62.5 μm MMF

Eye Patterns at fiber exit (1300 nm)

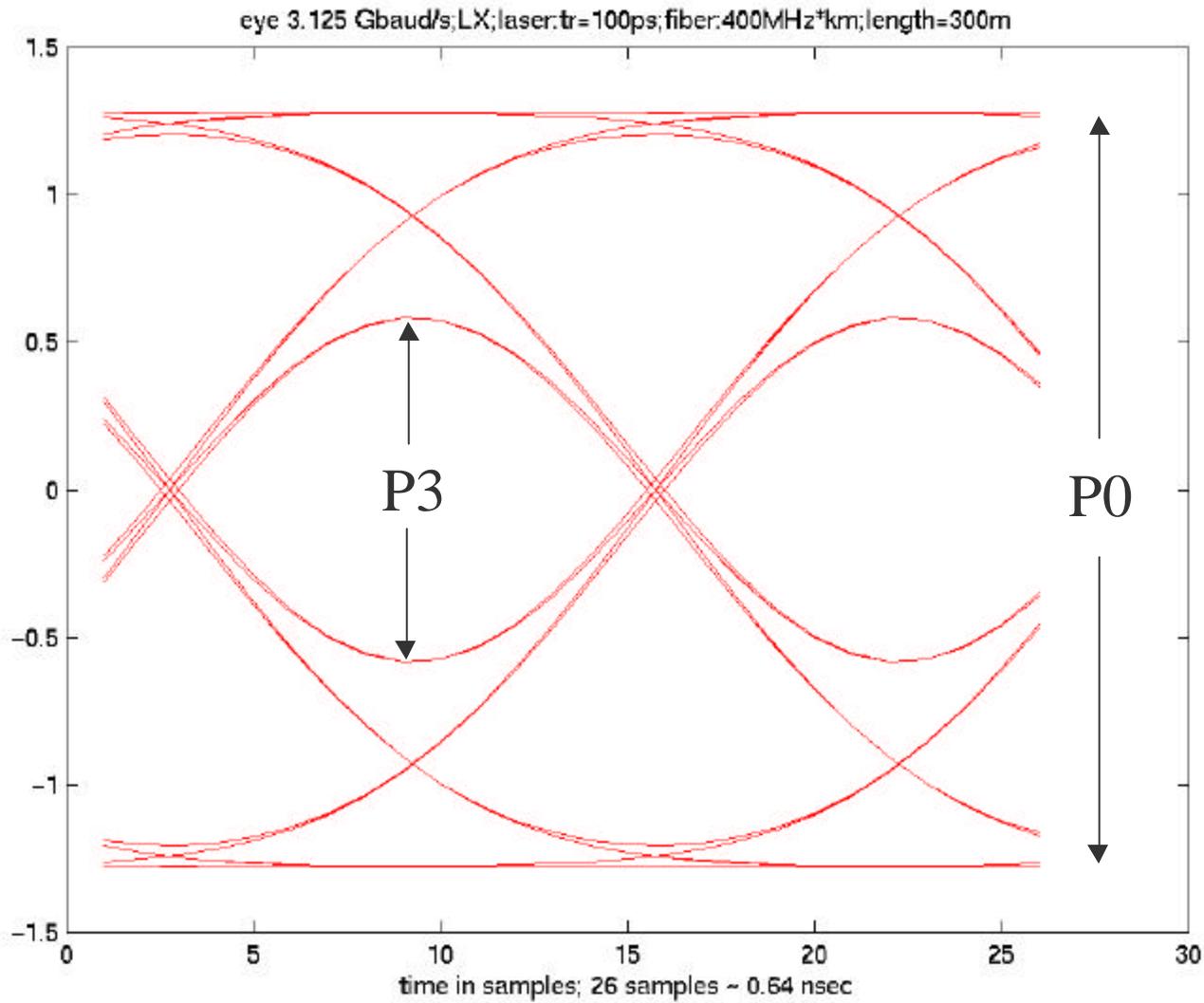
- Simulations use 1 GbE link model (see Ref 1) including laser rise-time, fiber modal and chromatic bandwidth, connector loss and cable attenuation at 1300 nm
- Simulated architectures:
 - PAM-5 + scrambling + 4-WDM @ 1.25 Gbaud/s
 - PAM-2 + 8b/10b + 4-WDM @ 3.125 Gbaud/s
 - PAM-5 + scrambling + serial @ 5 Gbaud/s

1.25 Gbaud/s - 300 meters



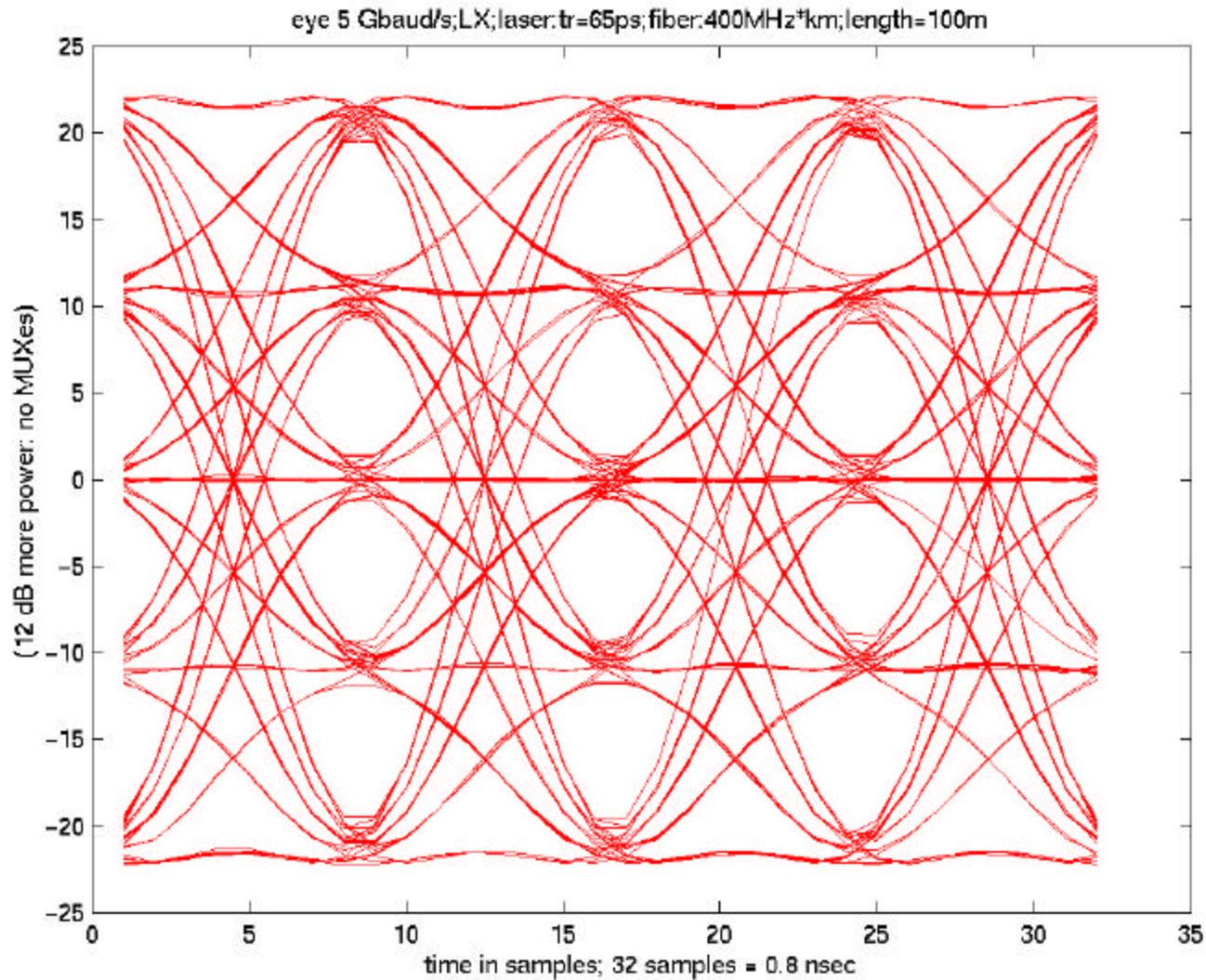
ISI loss = 0 dB

3.125 Gbaud/s - 300 meters



$$\text{ISI loss} = 10 \cdot \log(P0/P3) \sim 3.5 \text{ dB}$$

5 Gbaud/s - 100 meters (*)

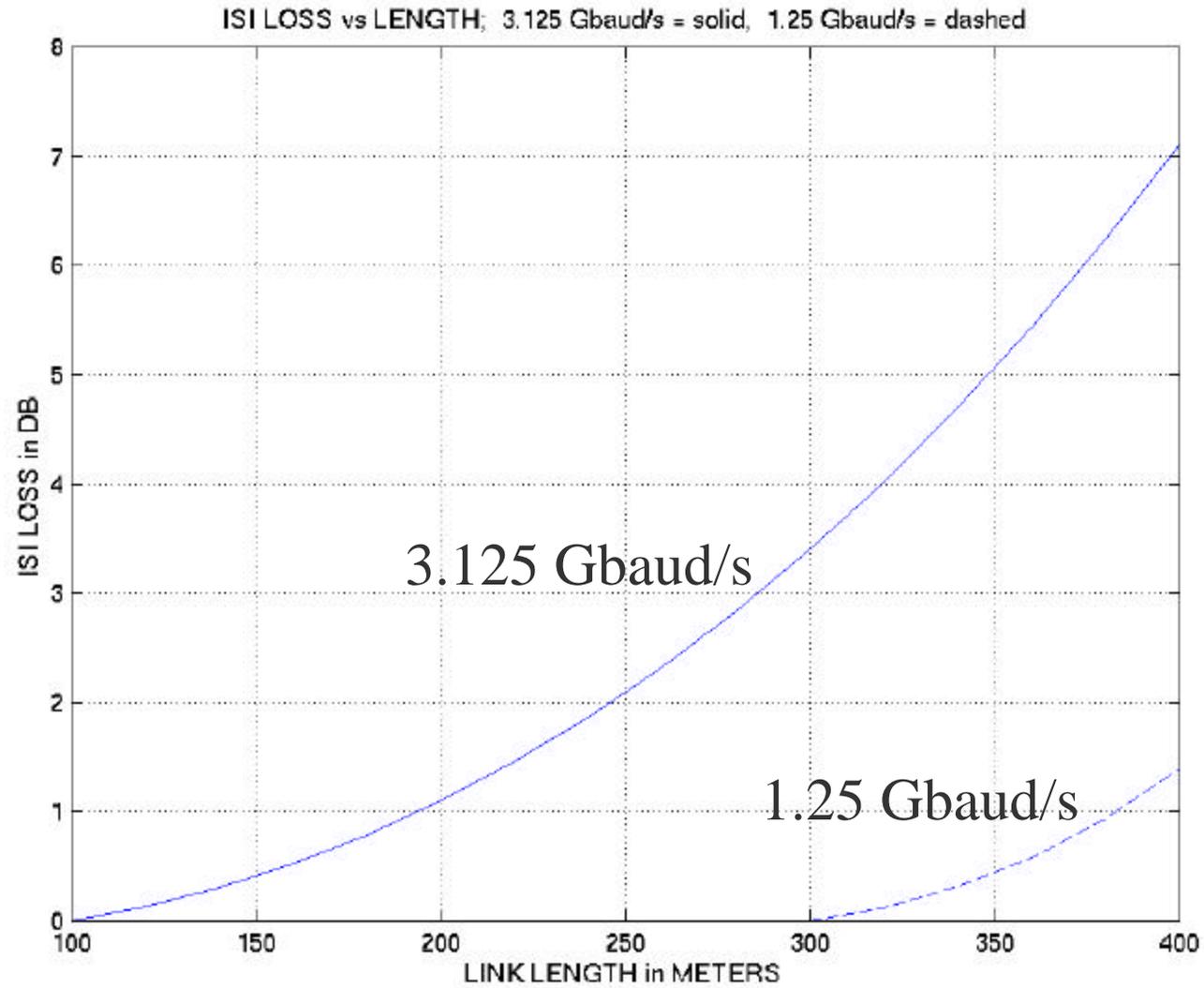


(*) eye closed at 200 meters, even with 500 MHz*km fiber

Eye patterns - 300m: Summary

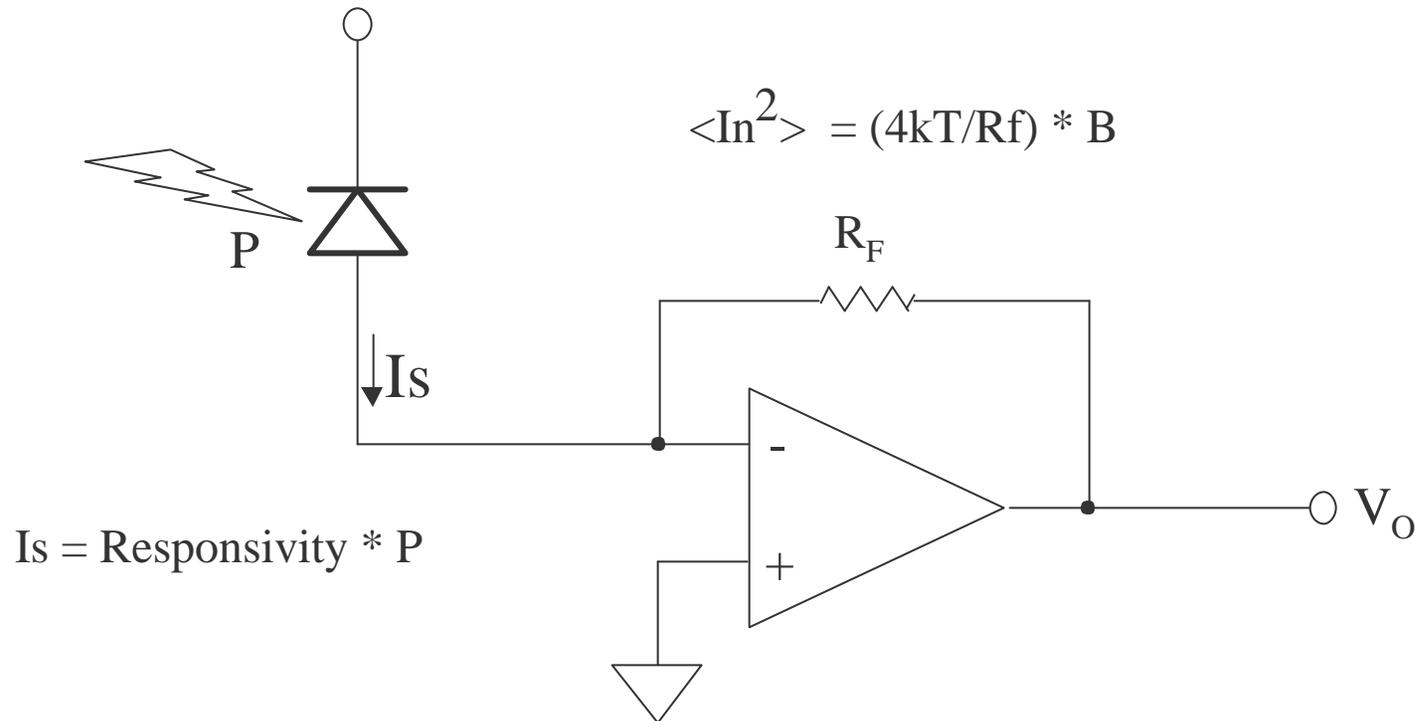
- ☞ ISI is negligible at 1.25 Gbaud/sec
- ☞ The system running at 3.125 Gbaud/sec has a large ISI penalty loss.
- ☞ The architecture at 5 Gbaud/s can not be used with 300 meters fiber (the eye is already closed at 200 meters)

ISI LOSS vs LINK LENGTH



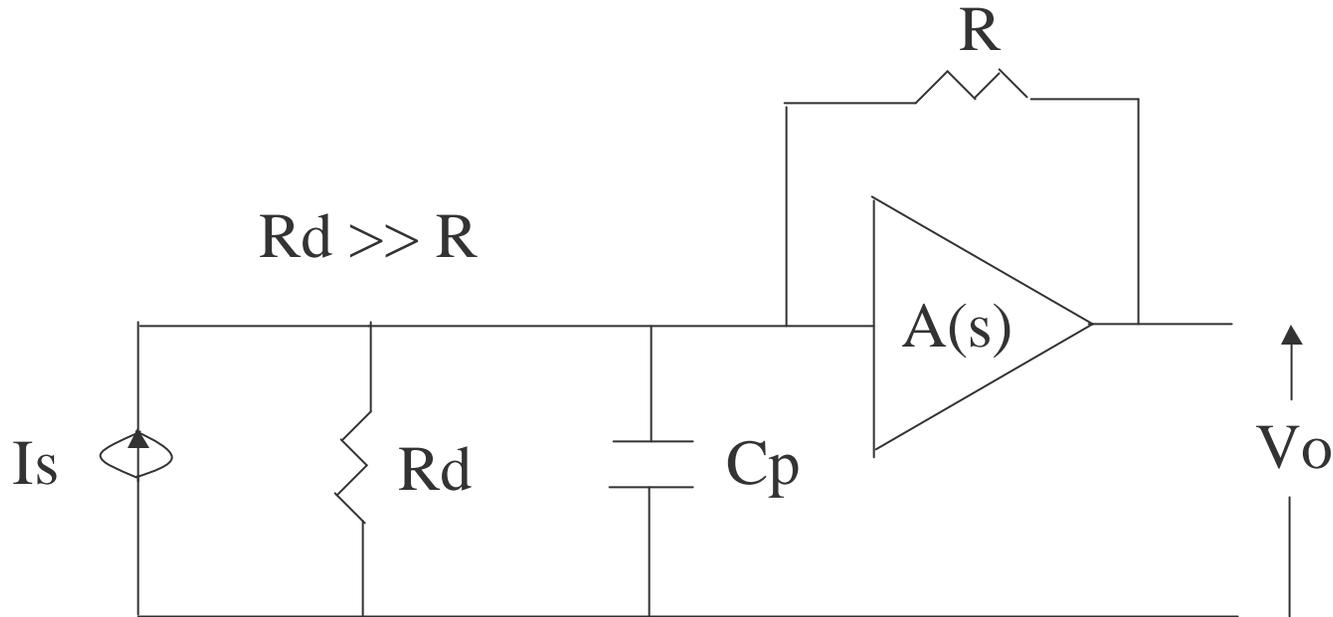
Receiver Front End

Assume PIN Photo Diode + Trans-Impedance Amplifier



$$\text{Electrical SNR} = 10 * \log (I_s^2 / I_n^2)$$

Small Signal Receiver Front End



$$A(s) = \frac{A_o}{1 + s/\omega_a}$$

$$\omega_p = \frac{1}{R * C_p}$$

Small Signal Transfer Function

$$V_o/I_s = -R * \frac{1}{1 + j*(1/Q)*(w/w_o) - (w/w_o)^2}$$

This is a 2nd order lowpass. $w_o = 2*\pi*B$. Usually, we set $Q = 1/\text{sqrt}(2)$ (Butterworth).

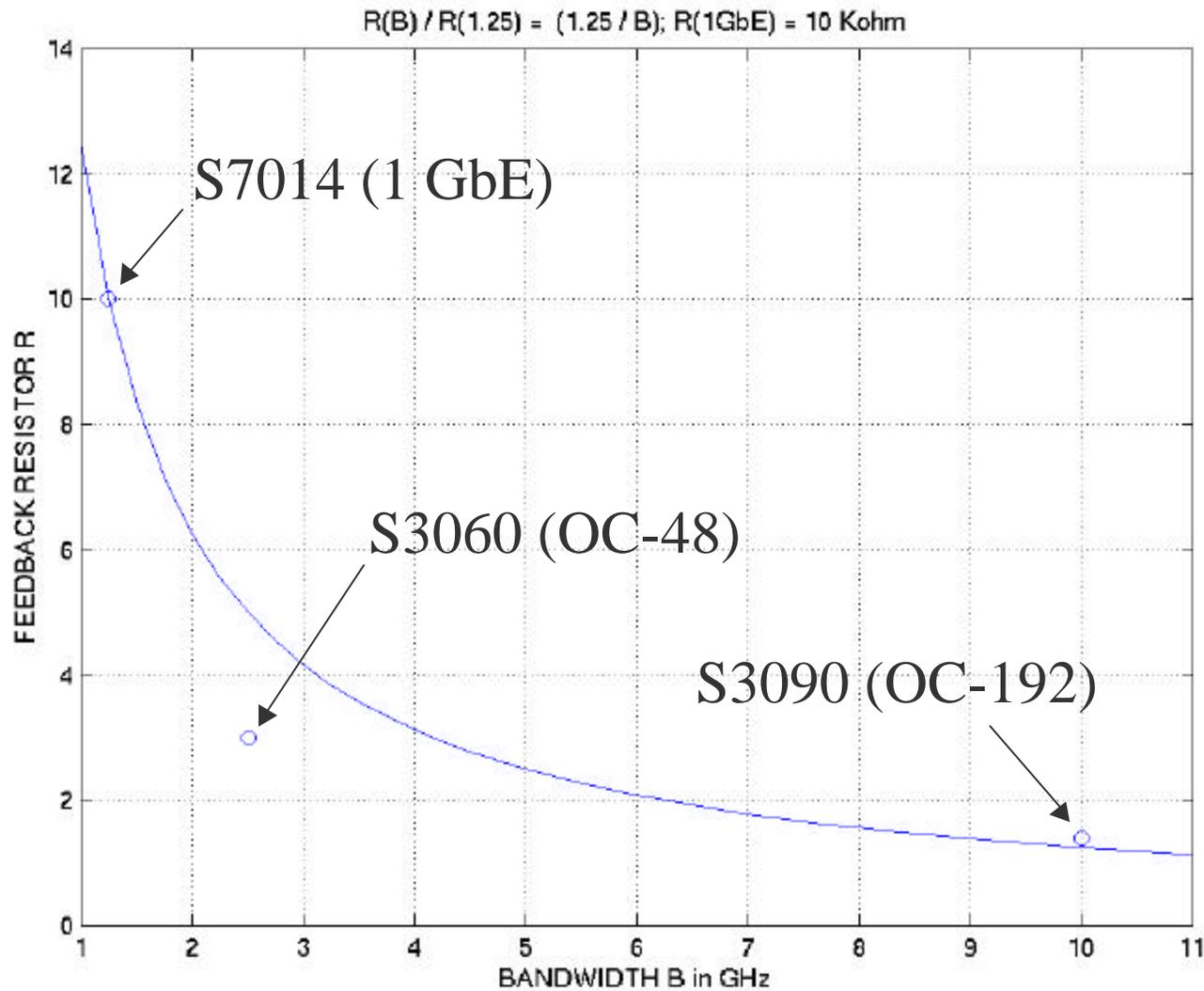
Assuming $A_o \gg 1$ and $w_a \ll w_o$ we obtain:

$$w_p = w_o/Q$$

or

$$R = \frac{1}{C_p} * \frac{Q}{w_o} = \frac{Q}{2*\pi*C_p} * \frac{1}{B}$$

Feedback Resistor vs Bandwidth



AMCC's Transimpedance Amplifiers

Thermal noise current

Replacing R into the equation for the thermal noise current, we finally obtain:

$$\langle I_n^2 \rangle = \frac{8 \cdot \pi \cdot k \cdot T \cdot C_p}{Q} * B^2$$

The thermal noise power is proportional to the **square** of the bandwidth B.

(for an alternative derivation see: Paul E. Green, “Fiber Optic Networks”, Prentice Hall, 1993, page 297, Eq 8.32)

Electrical SNR @ 300 m

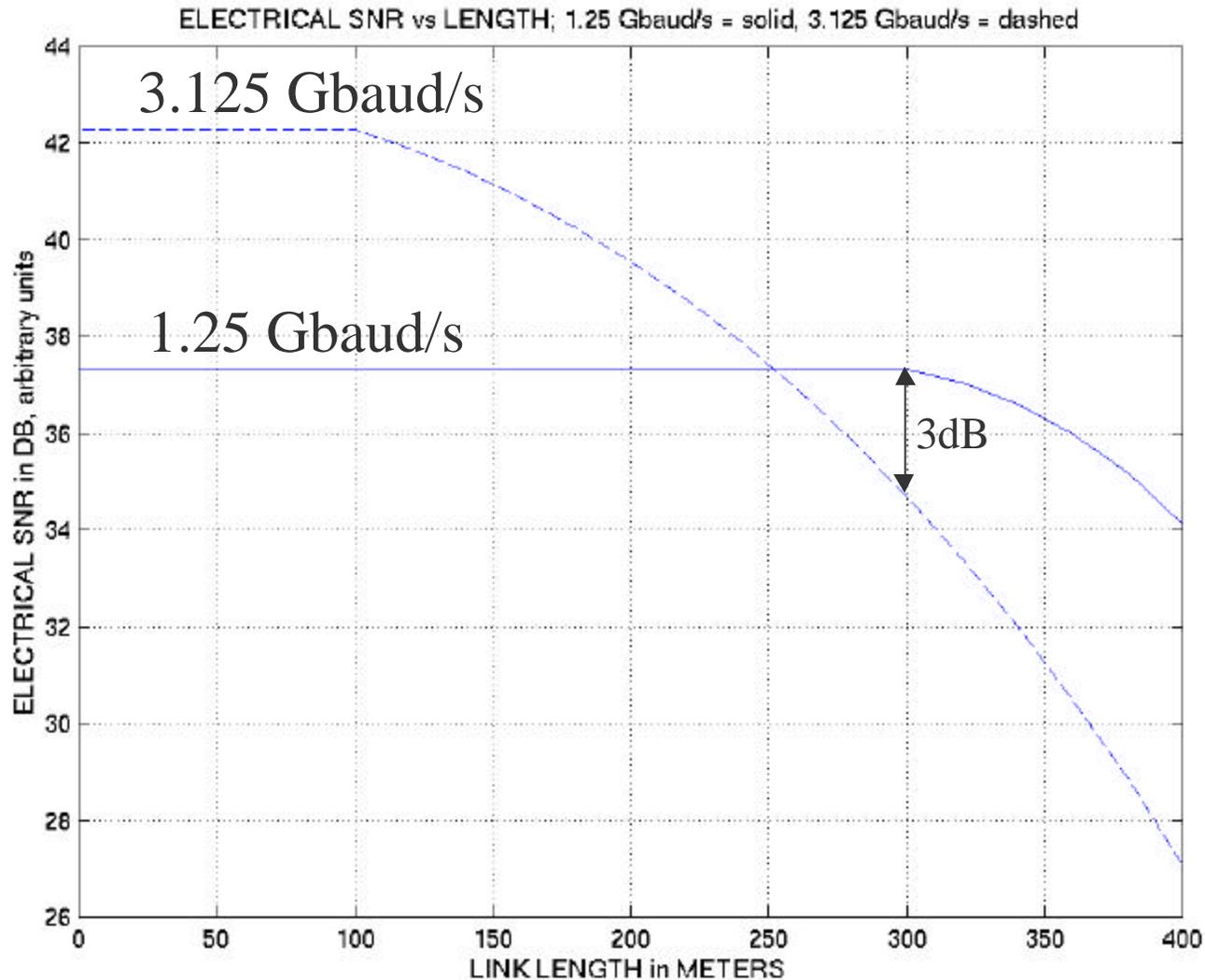
Neglecting any coding gain of the 1.25 Gbaud/s system, the relative SNRs @ 300 meters are:

$$\begin{aligned} \text{SNR}(1.25\text{Gb/s}) - \text{SNR}(3.125\text{Gb/s}) &= \\ &= 10 \cdot \log(P1/P3)^2 + 10 \cdot \log(B3/B1)^2 \end{aligned}$$

with $B3 = 3.125$, $B1 = 1.25$. Hence,

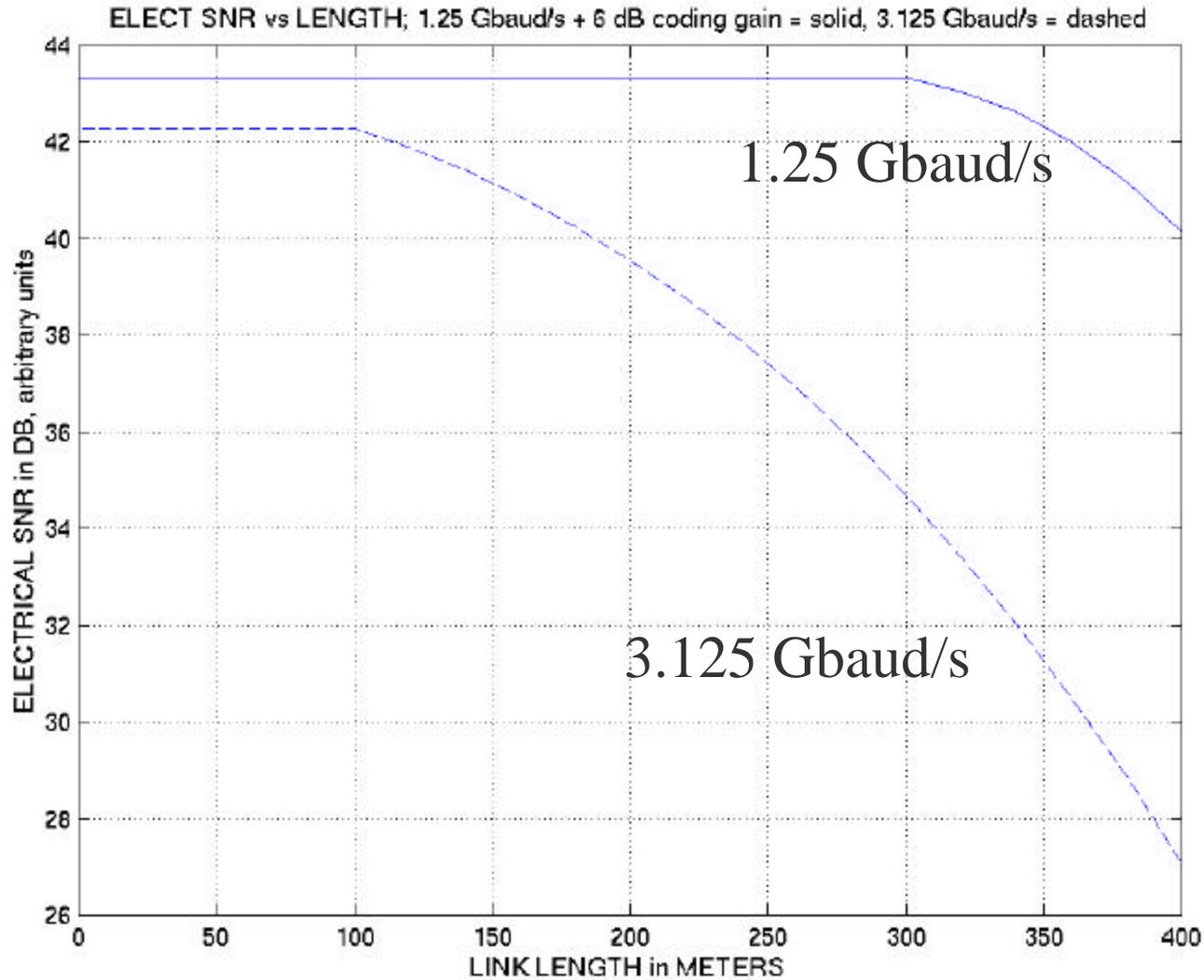
$$\text{SNR}(1.25\text{Gb/s}) - \text{SNR}(3.125\text{Gb/s}) \sim -5 + 8 = +3 \text{ dB}$$

ELECTRICAL SNR vs LINK LENGTH



(coding gain @ 1.25 Gbaud/s not included. It would shift the curve up)

ELECTRICAL SNR vs LINK LENGTH



(coding gain of 6 dB @ 1.25 Gbaud/s included)

Differential delay skew in 4-WDM

Use:

$$D = \frac{d}{d\lambda} \left(\frac{1}{V_g} \right) \quad (V_g = \text{group velocity})$$

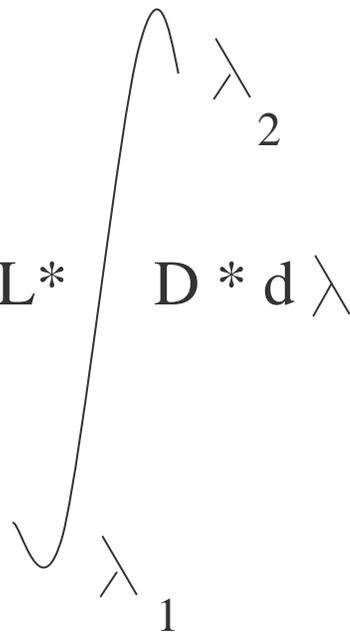
with D , a function of wavelength, given by Eq 9.10, Ref 1

Integrate:

$$D * d\lambda = d \left(\frac{1}{V_g} \right)$$

Differential delay skew

Differential delay between two wavelengths:

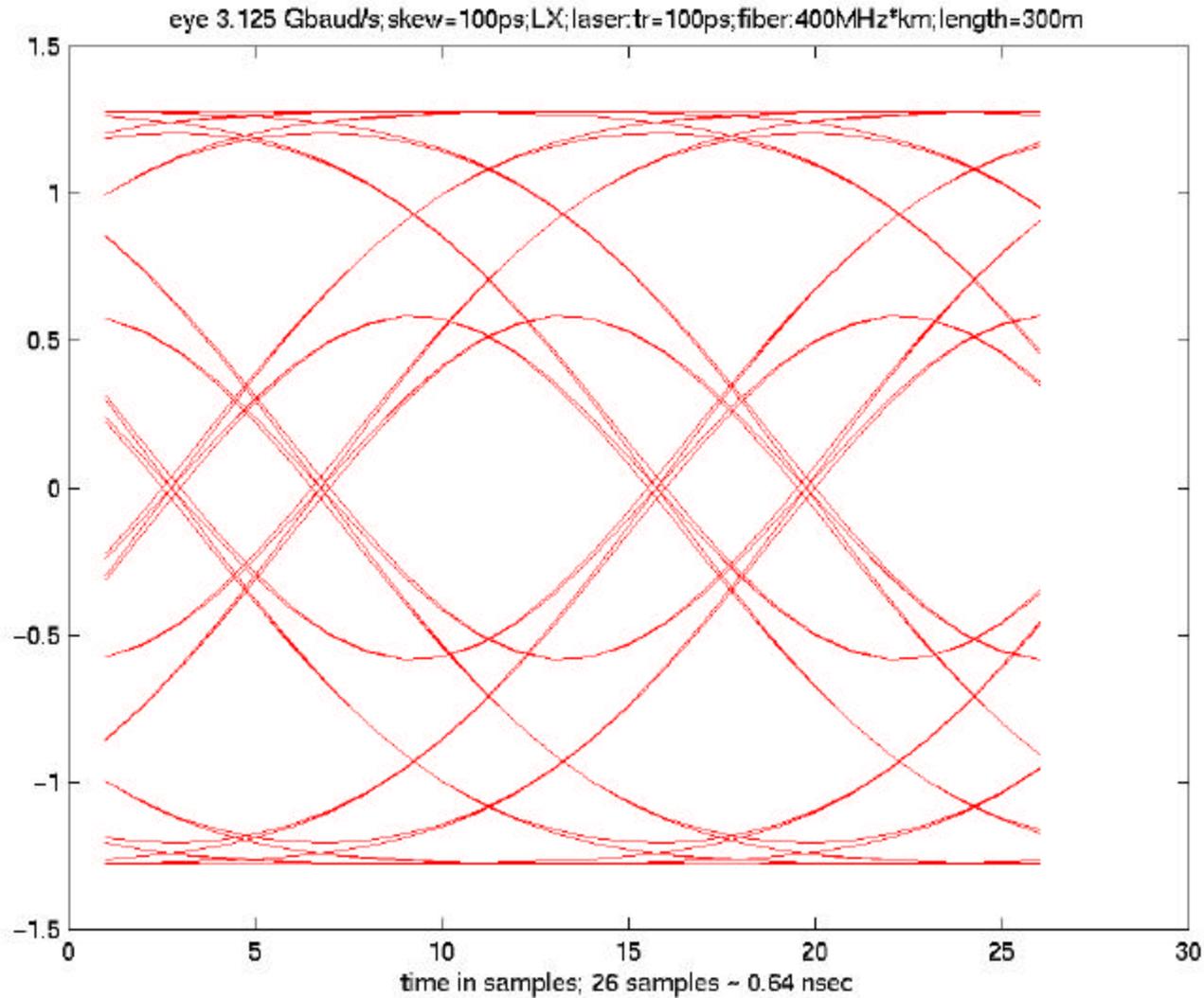

$$t_2 - t_1 = L * D * d\lambda$$

where L is the fiber length

Differential Delay Skew

- ✿ Using wavelengths of 1280,1300,1320 and 1340 nm (see Ref 2), the differential delay skew between the extreme wavelengths is given by:
 - $t_4 - t_1 \sim 0.33 * L$ (L in meters, time in psec)
- ✿ @ 300 m the differential delay skew is 100 psec.
- ✿ Assuming one clock recovery per Rx (same sampling phase for the 4 channels), the 3.125 Gbaud/s system (with only 320 psec baud period) will be more sensitive to the delay skew penalty.

3.125 Gbaud/s -300 meters - delay skew effects



(superimposed extreme wavelengths' eye patterns)

Summary of 3 architectures

PAM-5+ serial @ 5 Gbaud/s has a clear optical power advantage in shorter link lengths (no optical muxes), but ISI limits it to ~ 100 m link lengths.

8b/10b + 4-WDM @ 3.125 Gbaud/s provides a better coverage of the (0-200m) space, but at 300m ISI loss and delay skew penalty are large.

PAM-5 + 4-WDM @ 1.25 Gbaud/s becomes an attractive alternative, if used with coding gain, to provide the solution for the (0-300m) coverage space.

(continues in Part II)