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# 10GBASE-KR Transmitter Compliance Methodology Proposal

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# Scope and Purpose

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- Deficiencies of existing transmit template compliance methods are discussed.
- A proposal for a new compliance testing method capable of verifying a 10GBASE-KR transmitter is presented.
- Justification for tap range and tap resolution requirements will be presented separately.

# Agenda

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- Transmitter Compliance Testing Wish List
- Template Testing Discussion
- Equalization Ratio Testing
  - Transmitter model
  - Transmit Equalizer Signal Shaping
  - Transmit Equalizer Solution Space
  - Proposed Specification Methodology
- Conclusions

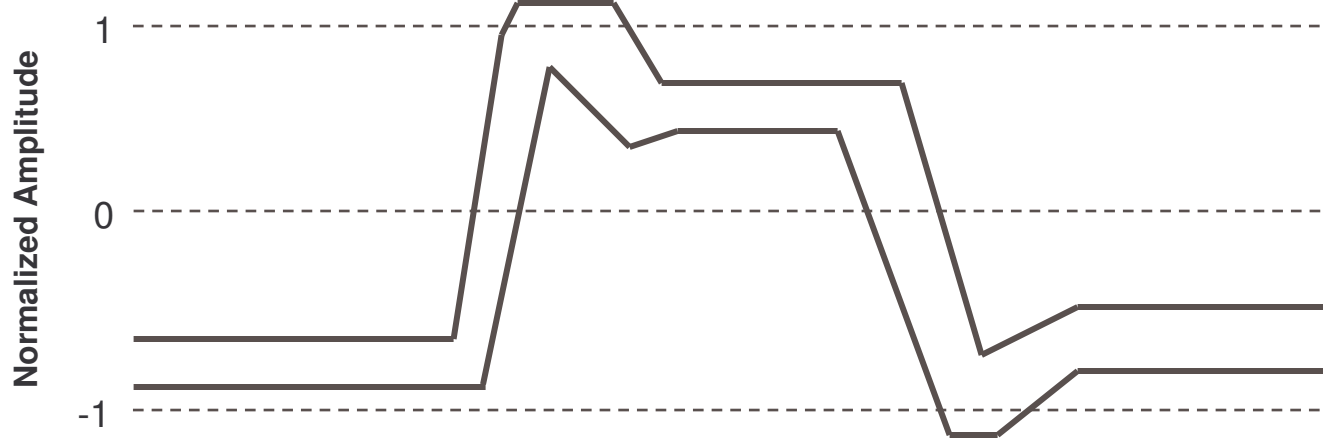
# Transmitter Compliance Testing Wish List

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- Simple
  - Easy to implement with existing test equipment
- Repeatable
  - Should test transmitter, not test equipment
- Fast
  - Minimize time required for compliance testing
- Definitive
  - Should clearly identify a compliant transmitter

# Template Testing Discussion

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- Bounds rise/fall times, jitter (to some extent), over-/undershoot and ringing, and relative amplitude.
- Bounds a single setting of pre- / de-emphasis (with tolerances).
- Documented as two voltage-time sequences (upper and lower bound).

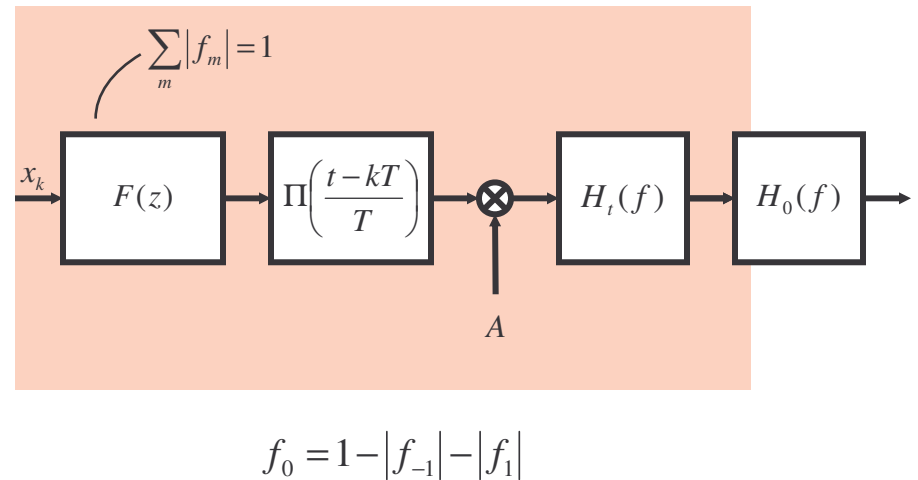
# Template Testing Discussion (continued)

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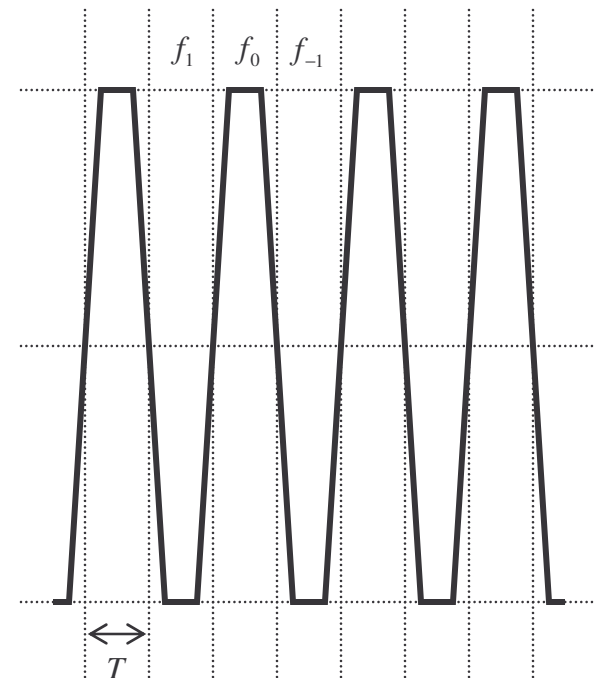
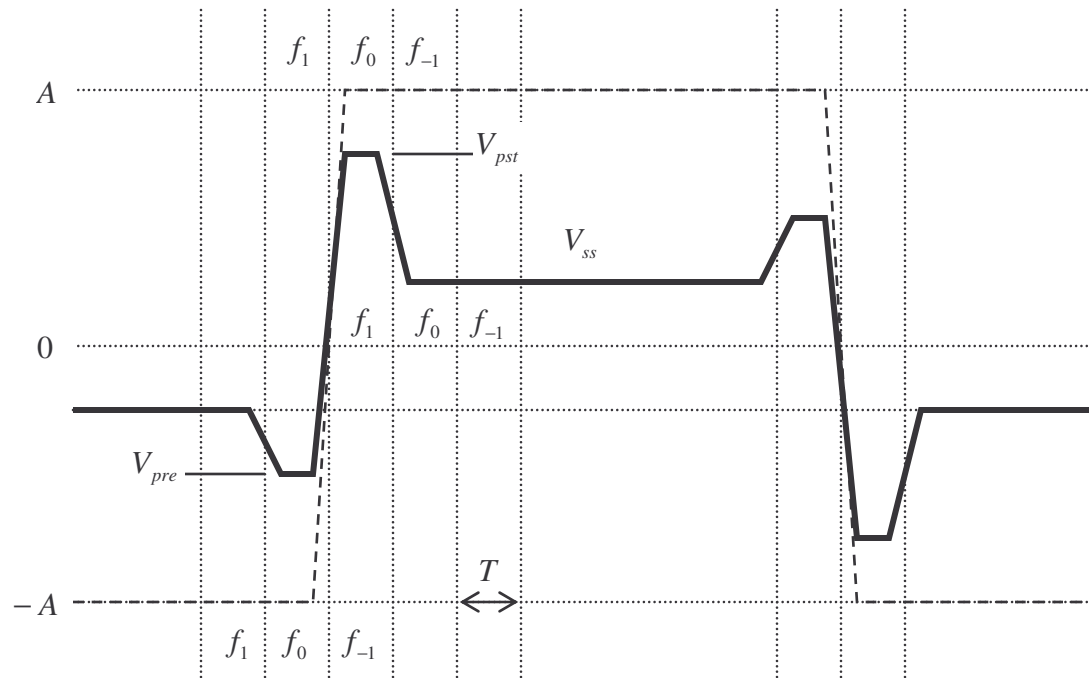
- Difficult to guarantee across all possible process, voltage, and temperature conditions
- Masks are generally relaxed to the point that they become meaningless
  - If mask is tight enough to guarantee a given level of equalization most designs will fail somewhere else
- Susceptible to instrumentation noise, calibration and capability
- Each template specifies only one equalization setting
- Variable transmit equalization will require multiple templates
  - Multiple templates will need to be used in characterization testing
  - Difficult to implement and document
- An algorithmic method may alleviate this problem.

# Transmitter Model

- $F(z)$  is the 3-tap transmit FIR
- $T$  is the symbol period
- $A$  is the peak differential output amplitude
- $H_t(f)$  is the transmit pulse shaping filter



# Transmit Equalizer Signal Shaping



$$V_{pre} = A(-f_1 - f_0 + f_{-1})$$

$$V_{pst} = A(-f_1 + f_0 + f_{-1})$$

$$V_{ss} = A(f_1 + f_0 + f_{-1})$$

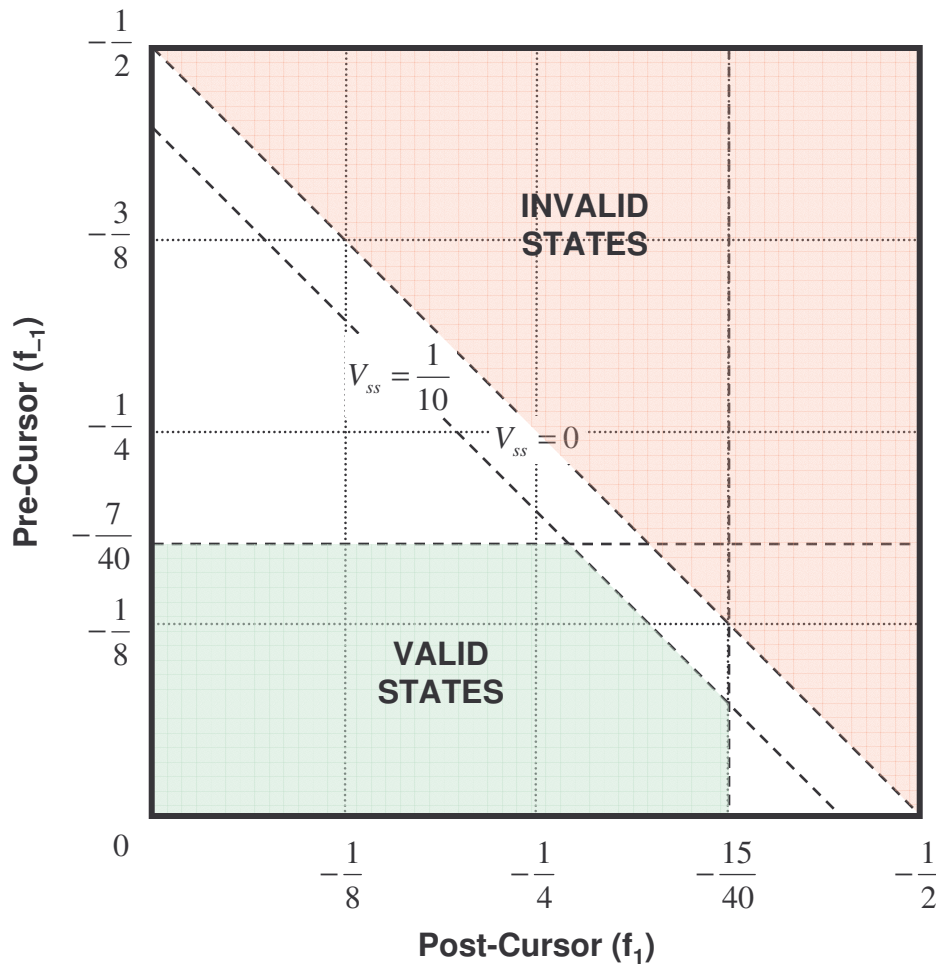
$$-f_1 + f_0 - f_{-1} = 1$$

**NOTE:**

By convention,  $f_1$  and  $f_{-1}$  are always negative and  $f_0$  is always positive.



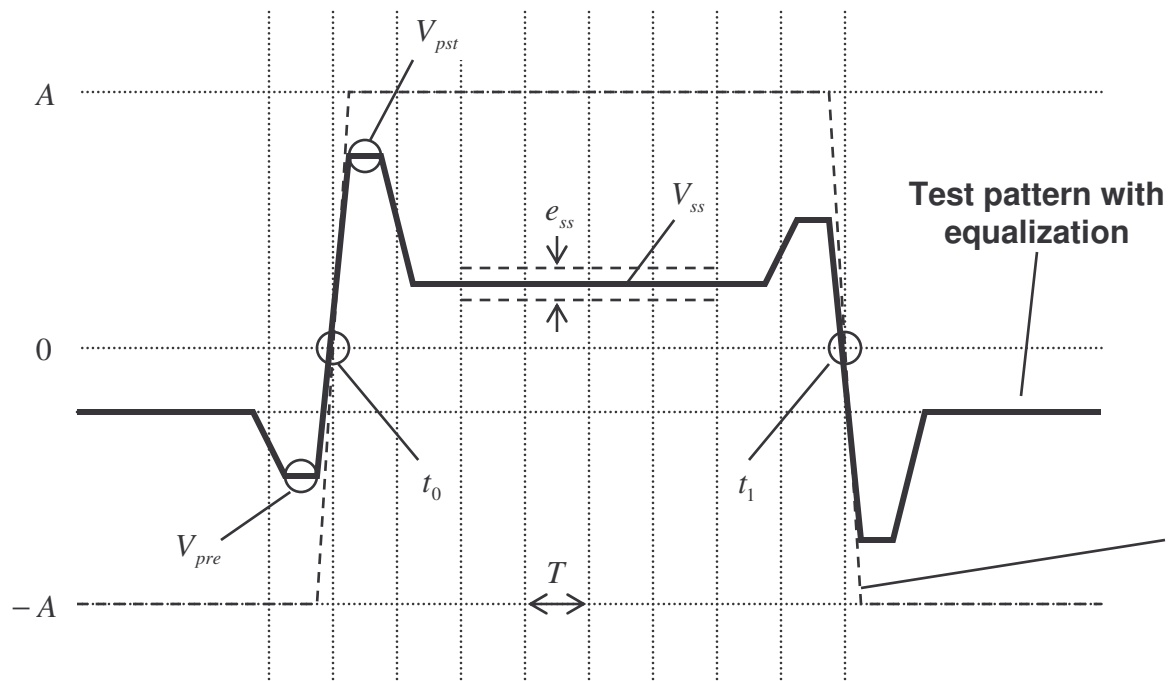
# Transmit Equalizer Solution Space



$f_{-1}$  : 8 steps from 0 to  $-\frac{7}{40}$   
 $f_1$  : 16 steps from 0 to  $-\frac{15}{40}$   
 $V_{ss}$  : minimum value should be no less than  $\frac{1}{10}$

VERIFY EACH TRANSMITTER STATE

# Proposed Specification Methodology



For each specified equalization setting, bound the following two ratios:

$$R_{pre} = \frac{V_{pre}}{-V_{ss}} \quad R_{pst} = \frac{V_{pre}}{V_{ss}}$$

Clause 49 square wave test pattern (recommend  $n \geq 8$ ), no equalization

## Definitions

- A** = peak transmit differential output amplitude
- T** = symbol period
- $t_0$**  = zero-crossing point of the rising edge of the AC-coupled signal
- $t_1$**  = zero-crossing point of the falling edge of the AC-coupled signal
- $V_{pre}$**  = peak voltage measured in the interval  $t_0 - T$  to  $t_0$
- $V_{pst}$**  = peak voltage measured in the interval  $t_0$  to  $t_0 + T$
- $V_{ss}$**  = steady-state voltage measured as the average voltage in the interval  $t_0 + 2T$  to  $t_1 - 2T$
- $e_{ss}$**  = steady-state error measured as the deviation from  $V_{ss}$  in the interval  $t_0 + 2T$  to  $t_1 - 2T$

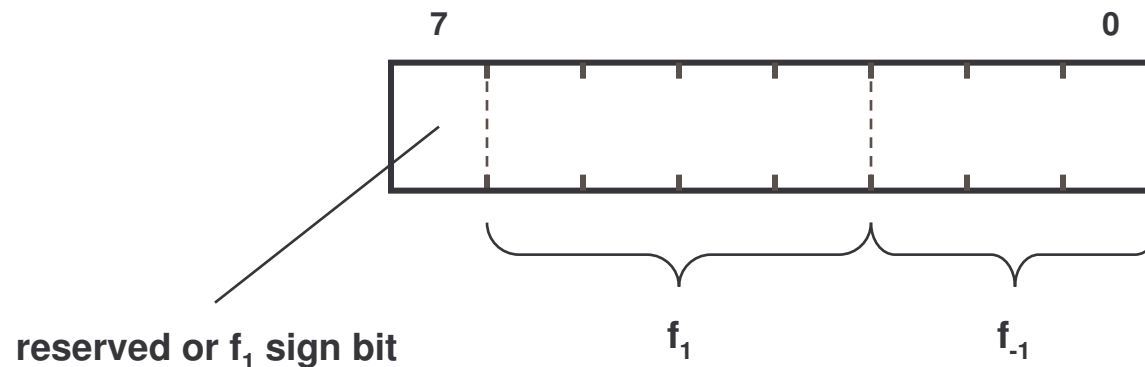
## Notes

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- Use “equalizer off” setting and clause 49 waveform to validate rise time, jitter, and amplitude requirements.
- May also want to check peak-peak output amplitude with 1010... pattern ( $n = 1$ ).

# Transmitter State Definition

- 8-bit encoding
- Pre- and post-cursor values encoded as  $N \times$  step size (for example,  $-0.025$ )



# Proposed Compliance Values

State <sub>10</sub>	R <sub>pre</sub> [dB]		R <sub>pst</sub> [dB]		R <sub>pre</sub> [V/V]		R <sub>pst</sub> [V/V]	
	(min)	(max)	(min)	(max)	(min)	(max)	(min)	(max)
0	-0.6	0.7	-0.6	0.7	0.93	1.08	0.93	1.08
8	-0.7	0.7	-0.2	1.1	0.93	1.08	0.98	1.14
16	-0.7	0.7	0.2	1.6	0.92	1.09	1.03	1.21
24	-0.8	0.8	0.7	2.2	0.92	1.09	1.08	1.28
32	-0.8	0.8	1.2	2.7	0.91	1.10	1.15	1.37
40	-0.9	0.9	1.7	3.3	0.91	1.11	1.22	1.46
48	-0.9	0.9	2.3	4.0	0.90	1.12	1.30	1.58
56	-1.0	1.0	2.9	4.7	0.89	1.13	1.39	1.71
64	-1.1	1.1	3.5	5.4	0.88	1.14	1.50	1.86
72	-1.2	1.2	4.2	6.2	0.88	1.15	1.63	2.05
80	-1.3	1.3	5.0	7.2	0.86	1.17	1.77	2.28
88	-1.4	1.5	5.8	8.2	0.85	1.19	1.95	2.56
96	-1.6	1.7	6.7	9.3	0.83	1.21	2.17	2.93
104	-1.8	1.9	7.7	10.7	0.81	1.25	2.44	3.42
112	-2.1	2.3	8.9	12.3	0.79	1.30	2.79	4.10
120	-2.5	2.8	10.2	14.2	0.75	1.38	3.25	5.13
1	-0.2	1.1	-0.7	0.7	0.98	1.14	0.93	1.08
⋮	⋮		⋮		⋮		⋮	

**Refer to spreadsheet  
for a complete list of  
values.**

# Conclusions

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- An algorithmic method for compliance testing the 10GBASE-KR transmit equalizer across settings compatible with the channels of interest is described.
- The described method has the advantages of being algorithmic and therefore easily implemented and documented.



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## Back-Up

# Tolerance Analysis

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$$V_{pre} = A(-f_1 - f_0 + f_{-1}) \longrightarrow V_{pre} = A(-1 - 2(f_1 + \Delta f_1))$$

$$V_{pst} = A(-f_1 + f_0 + f_{-1}) \longrightarrow V_{pst} = A(1 + 2(f_{-1} + \Delta f_{-1}))$$

$$V_{ss} = A(f_1 + f_0 + f_{-1}) \longrightarrow V_{ss} = A(1 + 2(f_1 + \Delta f_1) + 2(f_{-1} + \Delta f_{-1}))$$

$$\min(R_{pre}) = \frac{\max(V_{pre})}{\max(V_{ss})}$$

$$\max(R_{pst}) = \frac{\max(V_{pst})}{\min(V_{ss})}$$

$$\max(R_{pre}) = \frac{\min(V_{pre})}{\min(V_{ss})}$$

$$\min(R_{pst}) = \frac{\min(V_{pst})}{\max(V_{ss})}$$

$$\min(V_{pre}) = A(-1 - 2f_1 - 2|\Delta f_1|)$$

$$\min(V_{pst}) = A(1 + 2f_{-1} - 2|\Delta f_{-1}|)$$

$$\min(V_{ss}) = A(1 + 2(f_1 + f_{-1}) - 2(|\Delta f_1| + |\Delta f_{-1}|))$$

$$\max(V_{pre}) = A(-1 - 2f_1 + 2|\Delta f_1|)$$

$$\max(V_{pst}) = A(1 + 2f_{-1} + 2|\Delta f_{-1}|)$$

$$\max(V_{ss}) = A(1 + 2(f_1 + f_{-1}) + 2(|\Delta f_1| + |\Delta f_{-1}|))$$