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Power Supply Ripple and MDI Return Loss Modifications

GITESH BHAGWAT

ANALOG DEVICES



PSE and PD Power Supply Ripple

- ▶ PSE ripple voltage is measured using the test fixture shown here
- ▶ The Input impedance of the differential probe is given as:

$$Z_{in}(f) = (100 \pm 0.1\% \times \frac{\sqrt{f^2 + f_1^2}}{f})$$

- ▶ The Transfer function of the probe is given as:

$$H_1(f) = \frac{f}{\sqrt{f^2 + f_1^2}}$$

- This high pass filter emulates the high pass (PSE to PHY) effect of the power coupling network

- ▶ For ripple measured at the MDI, a $100\text{mV}_{\text{p-p}}$ limit is specified in Table 104-4 item 4a.

- ▶ For ripple seen at the PHY input, a $10\text{mV}_{\text{p-p}}$ limit is specified in Table 104-4 item 4b.

- To compare against this value, the measured ripple voltage is further post processed with the transfer given as:

$$H_2(f) = \frac{f}{\sqrt{f^2 + f_2^2}}$$

- This high pass filter emulates the high pass filter in the PHY

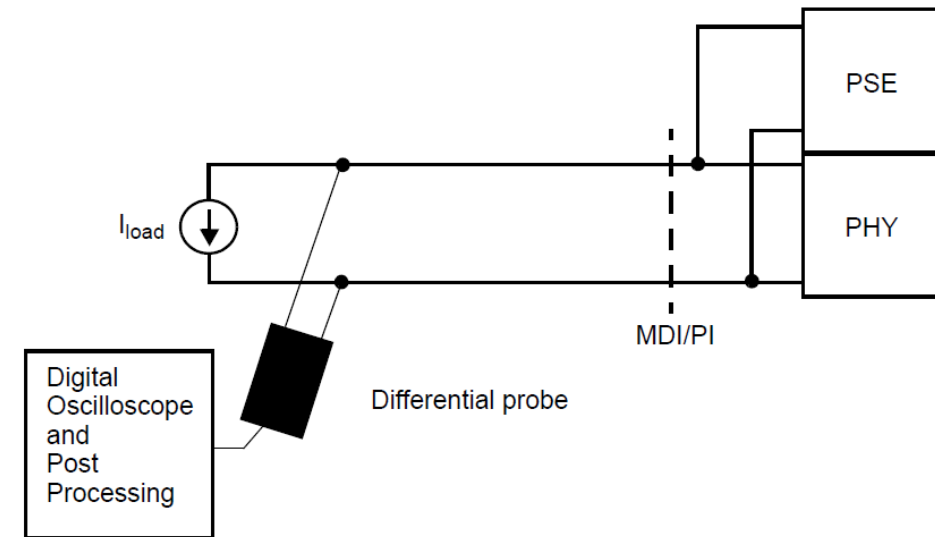


Figure 104-7—PSE ripple voltage test fixture

PSE and PD Power Supply Ripple

► The filter pole frequencies and the peak-to-peak ripple voltage values are shown below

PoDL Type	Data Speed	Modulation Scheme	Baud Rate	Ripple Filter Pole	
				f1	f2
Type E	10Mbps	PAM3	7.5 MBd	3.18 kHz \pm 1%	0.1 MHz \pm 1%
Type A, C	100Mbps	PAM3	66.66 MBd	31.8 kHz \pm 1%	1 MHz \pm 1%
Type B	1000Mbps	PAM3	750 MBd	318 kHz \pm 1%	10 MHz \pm 1%
Type F	2500Mbps	PAM4	1406.25 MBd	?	?
	5000Mbps		2812.5 MBd	?	?
	10000Mbps		5625 MBd	?	?

4	Power feeding ripple and noise:							
4a	1 kHz f <math>< 10</math> MHz		V_{p-p}	—	0.1	All	All	See 104.4.6.3
4b	1 kHz f <math>< 10</math> MHz			—	0.01	All	All	

PSE and PD Power Supply Ripple

- ▶ PAM4 (0.66V step) instead of PAM 3 (1V step)
 - Need more stringent PSE ripple specifications for NGAUTO systems
 - Scale peak ripple values from 0.1V to 0.066V for 4a and 0.01V to 0.0066V for 4b
- ▶ Coupling network in 1000BASE-T1 assumes a 3uH inductor and a 10nF capacitor ([gardner_3bu_2_0915.pdf](#))
 - Consider coupling network for NGAUTO with 2uH inductor and 10nF capacitor
 - Since ripple is measured at MDI, the HPF cutoff frequency determined by the RC pole and should remain same
 - $f_1 = 318\text{kHz}$
- ▶ Internal PHY filter cutoff – shift by baud rate
 - Consider 1406.25MBd for worst case filter. Scale PHY pole by a factor of 1.875 (compared against 750MBd)
 - $f_2 = 1.88 * 10\text{MHz} = 18.8\text{MHz}$
- ▶ Conclusion:
 - Lower peak to peak ripple voltage is allowed
 - The PHY filter has higher cutoff frequency
- ▶ Similar changes can be applied to PD ripple specifications

PSE Power Supply Ripple – Text Changes

- Change Table 104-4 to add the new ripple voltage levels for Type F PSEs as shown below:

Item	Parameter	Symbol	Unit	Min	Max	Class	Type	Additional Information
...
4	Power feeding ripple and noise:							
4a	1 kHz<f<10 MHz			-	0.1	All	All A,B,C,D,E	See 104.4.6.3
					0.066		F	
4b	1 kHz<f<10 MHz		V _{p-p}	-	0.01	All	All A,B,C,D,E	
					0.0066		F	
...

PSE Power Supply Ripple – Text Changes

- ▶ Change the edit to clause 104.4.6.3 to separate Type F and Type B PSEs and modify the cutoff frequencies:

(P62, L52) From:

“A digital oscilloscope or data acquisition module with a differential probe is used to observe the voltage at the MDI/PI of the PSE device under test (DUT) as shown in Figure 104–7. The input impedance, $Z_{in}(f)$, and transfer function, $H_1(f)$, of the differential probe are specified by Equation (104–1) and Equation (104–2), respectively. When measuring the ripple voltage for a Type A or Type C PSE as specified by Table 104–4 item (4a), $f_1 = 31.8 \text{ kHz} \pm 1\%$. When measuring the ripple voltage for a Type B or Type F PSE as specified in Table 104–4 item (4a), $f_1 = 318 \text{ kHz} \pm 1\%$.”

To:

A digital oscilloscope or data acquisition module with a differential probe is used to observe the voltage at the MDI/PI of the PSE device under test (DUT) as shown in Figure 104–7. The input impedance, $Z_{in}(f)$, and transfer function, $H_1(f)$, of the differential probe are specified by Equation (104–1) and Equation (104–2), respectively. When measuring the ripple voltage for a Type A or Type C PSE as specified by Table 104–4 item (4a), $f_1 = 31.8 \text{ kHz} \pm 1\%$. When measuring the ripple voltage for a Type B or Type F PSE as specified in Table 104–4 item (4a), $f_1 = 318 \text{ kHz} \pm 1\%$. When measuring the ripple voltage for a Type F PSE as specified in Table 104–4 item (4a), $f_1 = 318 \text{ kHz} \pm 1\%$.”

PSE Power Supply Ripple – Text Changes

- ▶ Change the edit to clause 104.4.6.3 to separate Type F and Type B PSEs and modify the cutoff frequencies:

(P63, L1) From:

“When measuring the ripple voltages for a Type B or Type F PSE as specified by Table 104–4 item (4b), the voltage observed at the MDI/PI with the differential probe where $f_1 = 318 \text{ kHz} \pm 1\%$ is post-processed with transfer function $H_2(f)$ specified in Equation (104–3) where $f_2 = 10 \text{ MHz} \pm 1\%$ ”

To:

“When measuring the ripple voltages for a Type B ~~or Type F~~ PSE as specified by Table 104–4 item (4b), the voltage observed at the MDI/PI with the differential probe where $f_1 = 318 \text{ kHz} \pm 1\%$ is post-processed with transfer function $H_2(f)$ specified in Equation (104–3) where $f_2 = 10 \text{ MHz} \pm 1\%$.

When measuring the ripple voltages for a Type F PSE as specified by Table 104–4 item (4b), the voltage observed at the MDI/PI with the differential probe where $f_1 = 318 \text{ kHz} \pm 1\%$ is post-processed with transfer function $H_2(f)$ specified in Equation (104–3) where $f_2 = 18.8 \text{ MHz} \pm 1\%$ ”

PD Power Supply Ripple – Text Changes

- Change Table 104-7 to add the new ripple voltage levels for Type F PDs as shown below:

Item	Parameter	Symbol	Unit	Min	Max	PD Type	Additional Information
...
3	Ripple voltage						
3a	1 kHz<f<10 MHz		V _{p-p}	-	0.1	A A,B,C,D,E	See 104.5.6.4
					0.066	F	
3b	1 kHz<f<10 MHz			-	0.01	A A,B,C,D,E	
					0.0066	F	
...

PD Power Supply Ripple – Text Changes

- ▶ Change the edit to clause 104.5.6.4 to separate Type F and Type B PDs and modify the cutoff frequencies:

(P63, L41) From:

“When measuring the ripple voltage for a Type B or Type F PD as specified by Table 104–7 item (3a), $f_1 = 318 \text{ kHz} \pm 1\%$.”

To:

“When measuring the ripple voltage for a Type B ~~or Type F~~ PD as specified by Table 104–7 item (3a), $f_1 = 318 \text{ kHz} \pm 1\%$.

When measuring the ripple voltage for a Type F PD as specified by Table 104–7 item (3a), $f_1 = 318 \text{ kHz} \pm 1\%$.”

(P63, L47) From:

“When measuring the ripple voltages for a Type B or Type F PD as specified by Table 104–7 item (3b), the voltage observed at the MDI/PI with the differential probe where $f_1 = 318 \text{ kHz} \pm 1\%$ shall be post-processed with transfer function $H_2(f)$ specified in Equation (104–3) where $f_2 = 10 \text{ MHz} \pm 1\%$.”

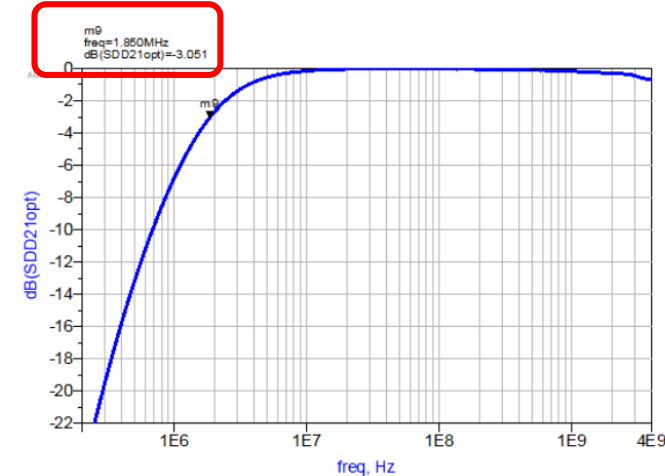
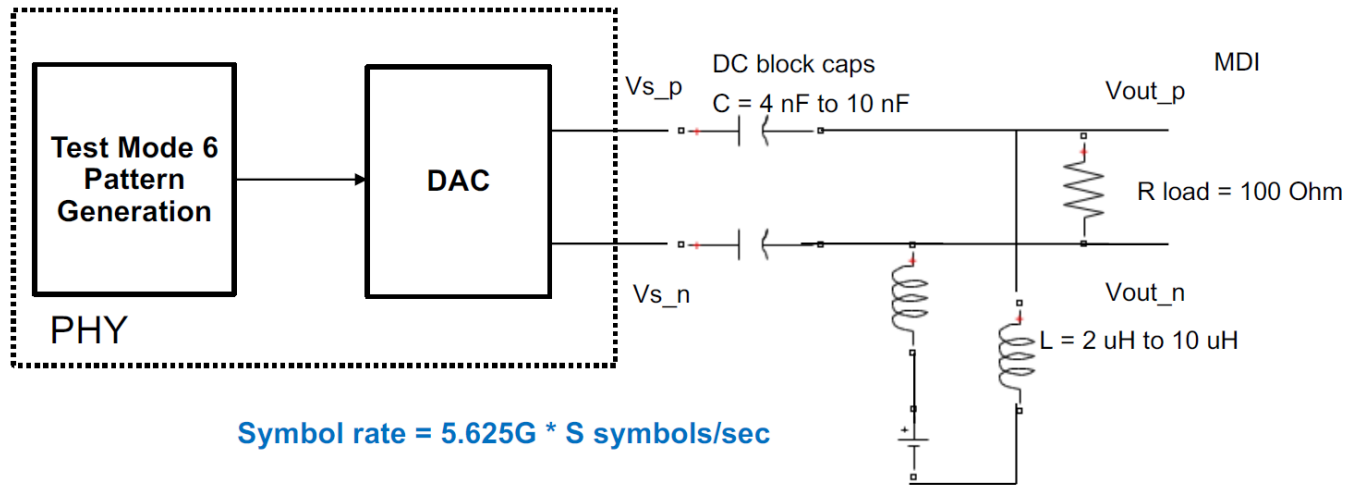
To:

“When measuring the ripple voltages for a Type B ~~or Type F~~ PD as specified by Table 104–7 item (3b), the voltage observed at the MDI/PI with the differential probe where $f_1 = 318 \text{ kHz} \pm 1\%$ shall be post-processed with transfer function $H_2(f)$ specified in Equation (104–3) where $f_2 = 10 \text{ MHz} \pm 1\%$.”

When measuring the ripple voltages for a Type F PD as specified by Table 104–7 item (3b), the voltage observed at the MDI/PI with the differential probe where $f_1 = 318 \text{ kHz} \pm 1\%$ shall be post-processed with transfer function $H_2(f)$ specified in Equation (104–3) where $f_2 = 18.8 \text{ MHz} \pm 1\%$.”

Low Frequency MDI Return Loss and Transmitter Droop

- ▶ Transmitter droop was specified considering a 2uH inductance and 10nF capacitance per transmitter output ([souvignier_3ch_02_0319.pdf](#))
- ▶ This yields an insertion loss 3dB HPF pole at 1.85MHz
- ▶ Need to adjust MDI return loss mask to align with the coupling network used for droop
 - Same Insertion loss 3dB pole

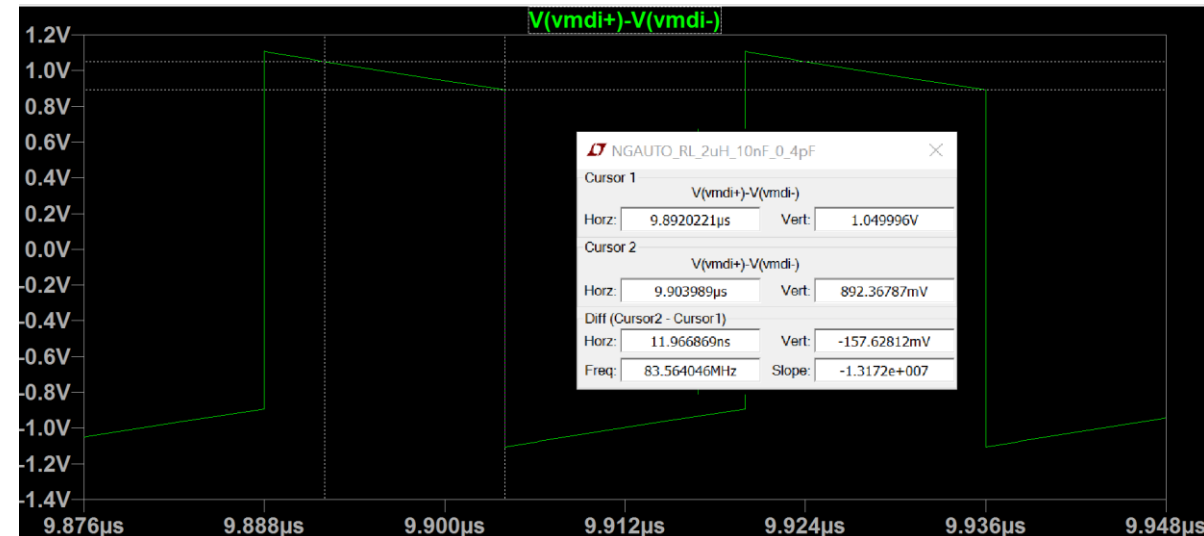
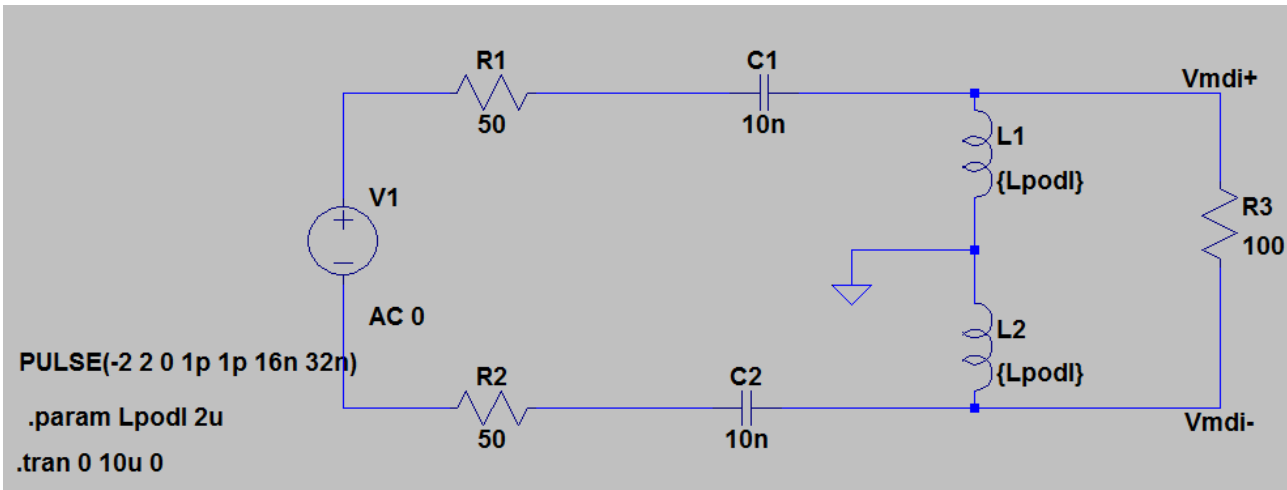


DC cap = 10 nF
 L = 2 uH , 0.15 pF internal parasitic cap
 R load = 100 Ohm
 R source = 100 Ohm
 Board Trace = 1 inch
 Connector = Rosenberger H-MTD

Droop = 15 %
 HighPass_3dB = 1.85 MHz

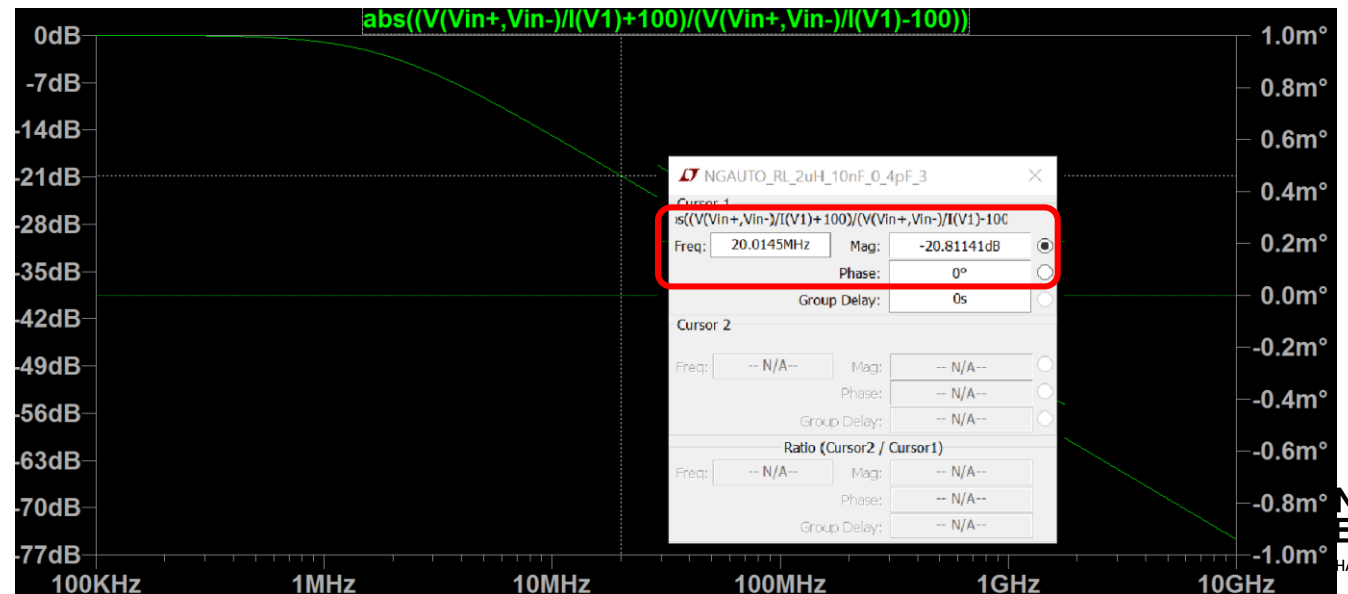
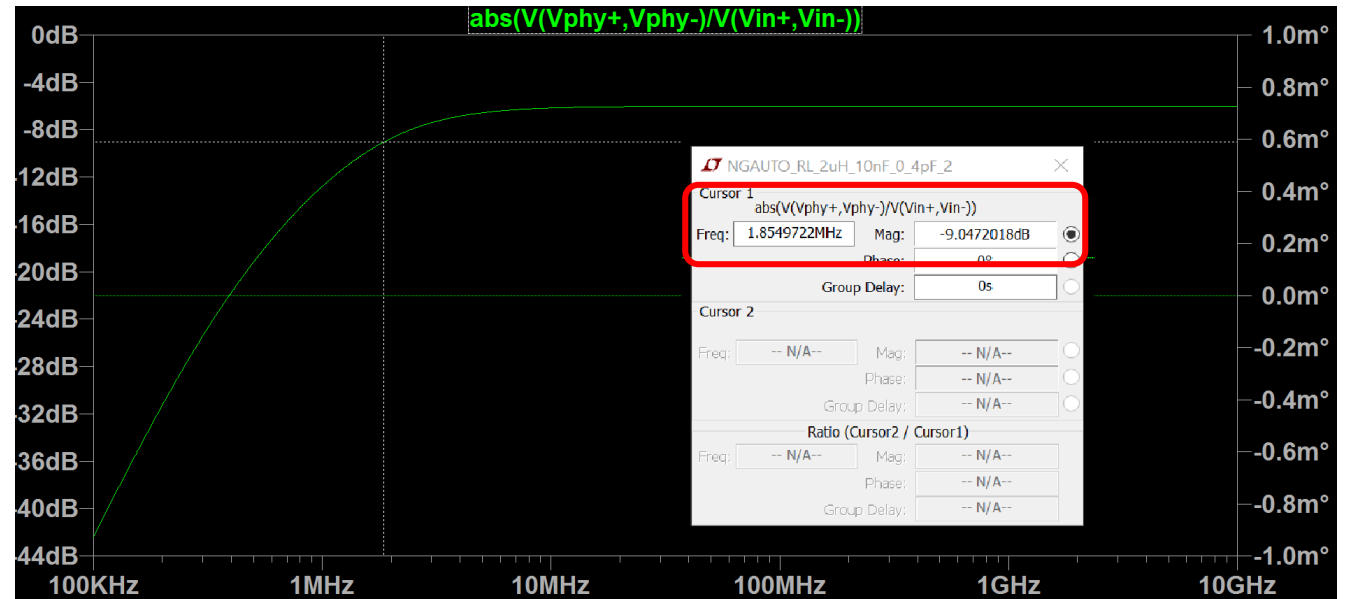
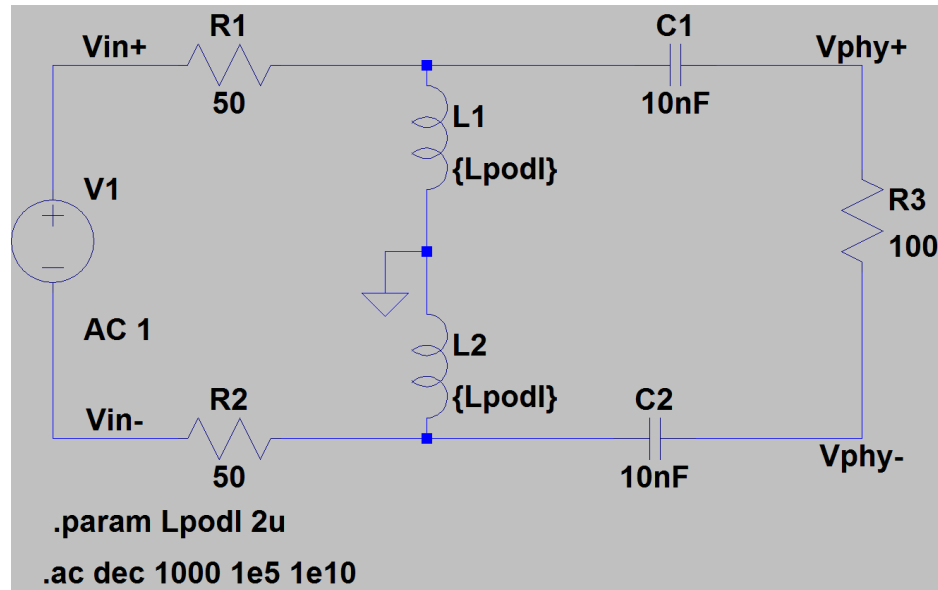
Low Frequency MDI Return Loss and Transmitter Droop

- ▶ Coupling circuit for droop simulation:
 - 2uH coupling inductor
 - 10nF coupling capacitor
- ▶ Droop calculated = 15%
 - $\text{Droop} = \frac{(1.0499 - 0.89236)V}{1.0499V} \times 100$ in 12ns
- ▶ Verify insertion loss HPF pole at 1.85Mhz
- ▶ Measure low frequency return loss for these values



Low Frequency MDI Return Loss and Transmitter Droop

- ▶ HPF pole in insertion loss at 1.85 MHz verified
- ▶ Return Loss has a breakpoint of
 - -20dB at 20MHz
- ▶ Change the low frequency MDI return loss breakpoint to align with this



Low Frequency MDI Return Loss Text Changes

- ▶ Change the edit to clause 149.8.2.1 MDI return loss to change the low frequency breakpoint

- ▶ **(P168, L2)** From:

$$MDI_Return_Loss(f) \leq \left\{ \begin{array}{ll} 20 - 20 \left(\log_{10} \frac{10}{f} \right) & 1 \leq f < 10 \\ 20 & 10 \leq f \leq 500 \\ 12 - 10 \log_{10}(f/3000) & 500 \leq f \leq 3000 \\ 12 - 20 \log_{10}(f/3000) & 3000 \leq f \leq 4000 \end{array} \right\} \text{ (dB)} \quad (149-27)$$

where

f is the frequency in MHz.

- ▶ To:

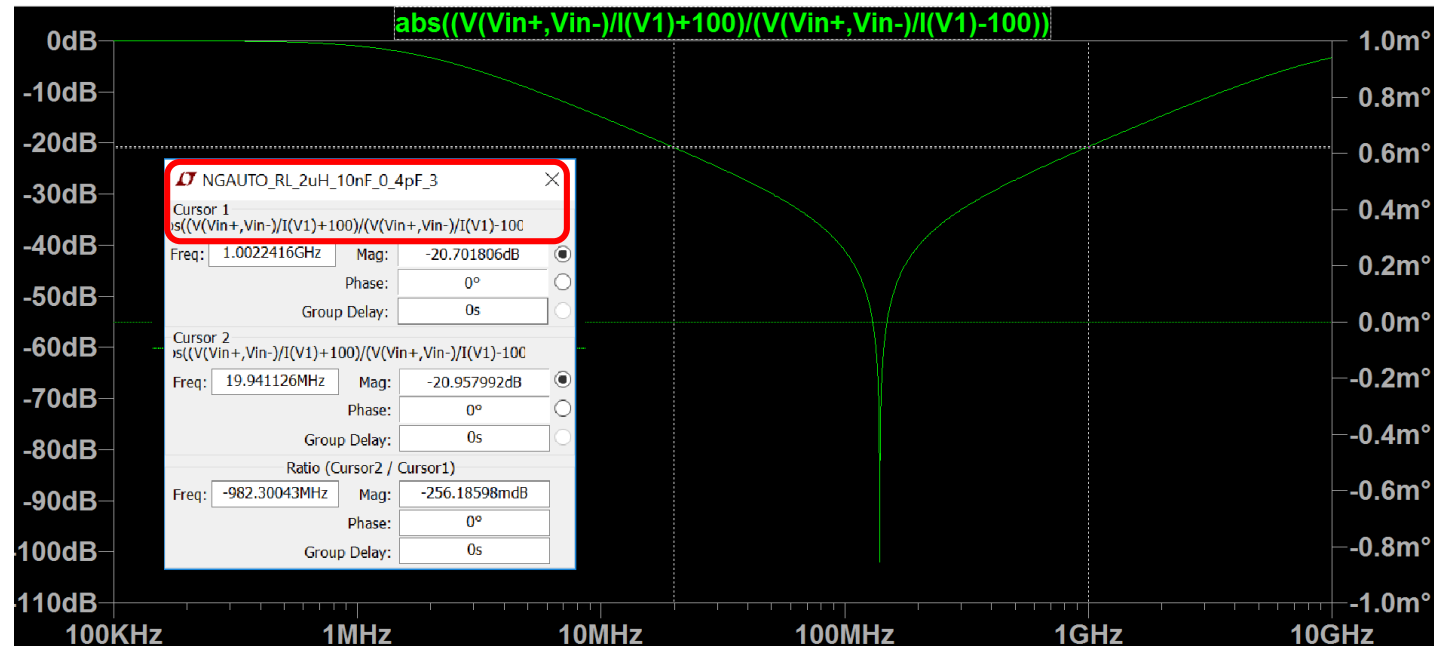
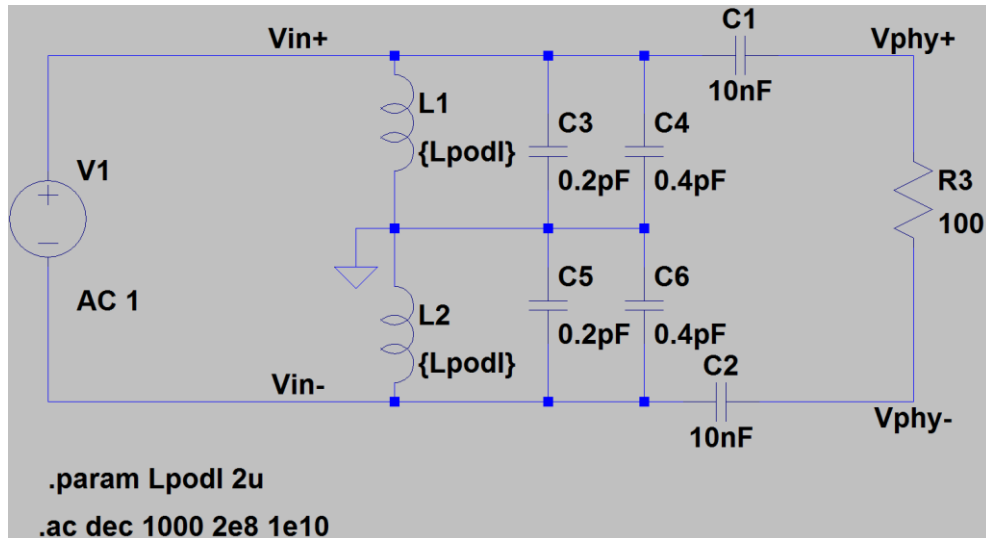
Return Loss \geq

- **$20 - 20 \times \text{Log}_{10}\left(\frac{20}{f}\right)$** **for $2 \leq f \leq 20$**
- **20** **for $20 \leq f \leq 500$**
- $12 - 10 \times \text{Log}_{10}\left(\frac{f}{3000}\right)$ for $500 \leq f \leq 3000$
- $12 - 20 \times \text{Log}_{10}\left(\frac{f}{3000}\right)$ for $3000 \leq f \leq 4000$

where f is frequency in MHz

High Frequency MDI Return Loss and ESD Protection Devices

- ▶ PHY devices may need additional protection using devices such as ESD clamping diodes
 - Consider additional capacitive loading of 0.4pF per output
- ▶ Coupling inductors have parasitic capacitance
 - Considering an SRF of about 250MHz, capacitance of 0.2pF per inductor
- ▶ This yields a high frequency return loss of -20dB at 1GHz
- ▶ Adding further margin for termination tolerance, trace inductance, package inductances etc.
 - ▶ Consider a breakpoint of -20dB at 500MHz
 - ▶ And a return loss of -5dB at 4000MHz



High Frequency MDI Return Loss Text Changes

- ▶ Change the edit to clause 149.8.2.1 MDI return loss to change the high frequency mask

- ▶ **(P168, L2)** From:

$$MDI_Return_Loss(f) \leq \left\{ \begin{array}{ll} 20 - 20 \left(\log_{10} \frac{10}{f} \right) & 1 \leq f < 10 \\ 20 & 10 \leq f \leq 500 \\ 12 - 10 \log_{10}(f/3000) & 500 \leq f \leq 3000 \\ 12 - 20 \log_{10}(f/3000) & 3000 \leq f \leq 4000 \end{array} \right\} \text{ (dB)} \quad (149-27)$$

where

f is the frequency in MHz.

- ▶ To:

Return Loss \geq

- $20 - 20 \times \text{Log}_{10}\left(\frac{10}{f}\right)$ for $1 \leq f \leq 10$
- 20 for $10 \leq f \leq 500$
- **$20 - 16.5 \times \text{Log}_{10}\left(\frac{f}{500}\right)$ for $500 \leq f \leq 4000$**

where f is frequency in MHz

All MDI Return Loss Text Changes and Comparison

Existing mask:

$$MDI_Return_Loss(f) \leq \begin{cases} 20 - 20 \left(\log_{10} \frac{10}{f} \right) & 1 \leq f < 10 \\ 20 & 10 \leq f \leq 500 \\ 12 - 10 \log_{10}(f/3000) & 500 \leq f \leq 3000 \\ 12 - 20 \log_{10}(f/3000) & 3000 \leq f \leq 4000 \end{cases} \text{ (dB)}$$

where

f is the frequency in MHz.

Modified mask:

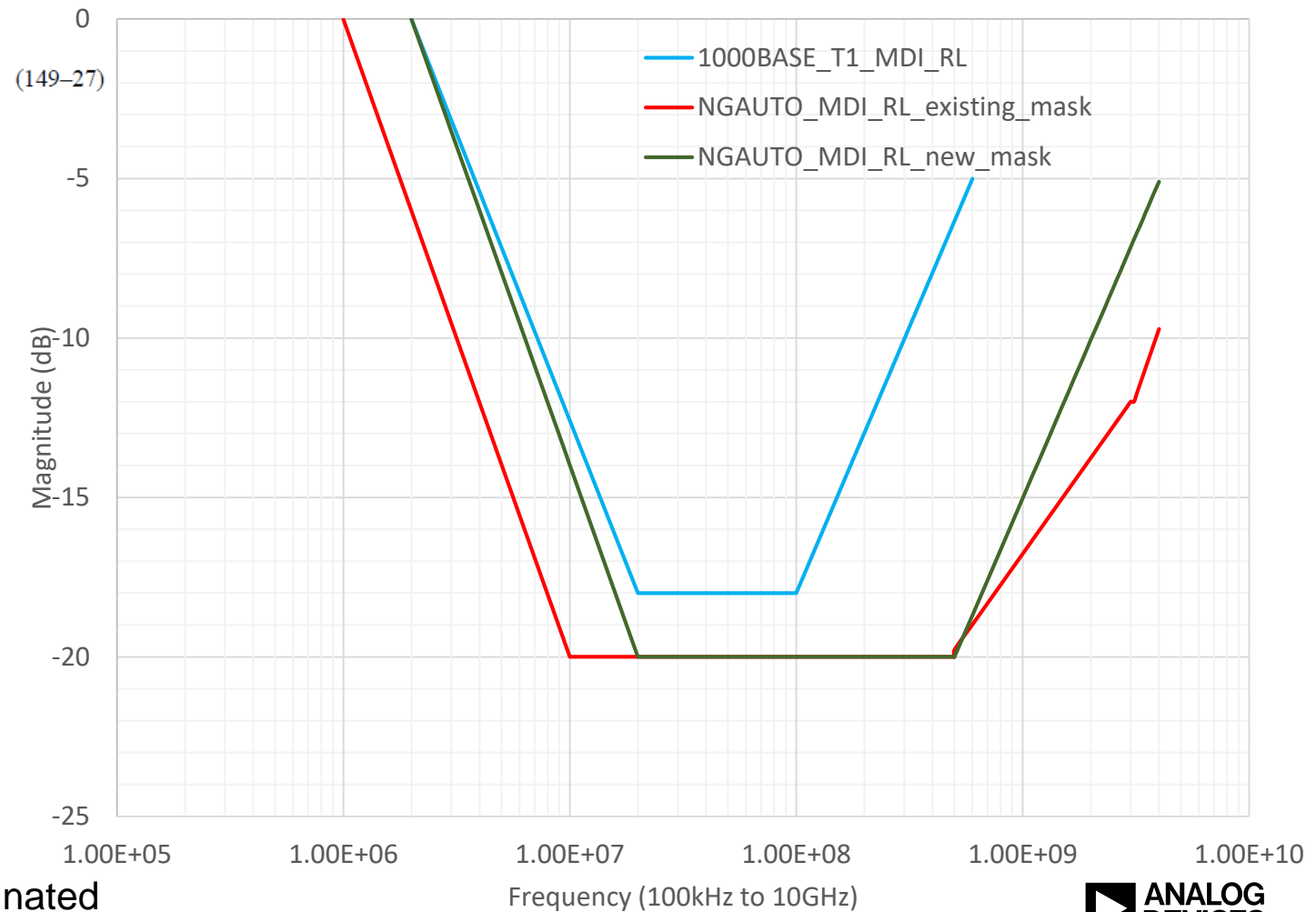
Return Loss \geq

- $20 - 20 \times \text{Log}_{10}\left(\frac{20}{f}\right)$ for $2 \leq f \leq 20$
- 20 for $20 \leq f \leq 500$
- $20 - 16.5 \times \text{Log}_{10}\left(\frac{f}{500}\right)$ for $500 \leq f \leq 4000$

where f is frequency in MHz

► Note: discontinuity in previous mask has been eliminated

MDI Return Loss Masks- Comparison



Thank You!

QUESTIONS? FEEDBACK?

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