Project	IEEE 802.16 Broadband Wireless Access Working Group < <u>http://ieee802.org/16</u> >
Title	LDPC coding for OFDMA PHY
Date Subm itted	2004-05-01
Source(s)	Panyuh Joo, Seho Myung, Jaeyeol Ki m, Gyubum Kyung, Hongsil Jeong, K yungcheol Yang , DS Park, Jeho JeonVoice: +82-31-279-5096 Fax: +82-31-279-5130 
Re:	Reply comments for the sponsor re-circulation Ballot
Abstract	Enhanced LDPC coding scheme
Purpose	This document id for discussion with INTEL
Notice	This document has been prepared to assist IEEE 802.16. It is offered as a basis for discussion and is not binding on the contributing individual(s) or organization(s). The material in this document is subject to change in form and con tent after further study. The contributor(s) reserve(s) the right to add, amend or withdraw material contained herein. The contributor grants a free, irrevocable license to the IEEE to incorporate material contained in this contribution,
Release	and any modifications thereof, in the creation of an IEEE Standards publication; to copyright in the IEEE's name an y IEEE Standards publication even though it may include portions of this contribution; and at the IEEE's sole discr etion to permit others to reproduce in whole or in part the resulting IEEE Standards publication. The contributor als o acknowledges and accepts that this contribution may be made public by IEEE 802.16.
Patent Poli cy and Proc edures	The contributor is familiar with the IEEE 802.16 Patent Policy and Procedures (Version 1.0) < <u>http://ieee802.org/16</u> /ipr/patents/policy.html>, including the statement "IEEE standards may include the known use of patent(s), includin g patent applications, if there is technical justification in the opinion of the standards- developing committee and provided the IEEE receives assurance from the patent holder that it will license applicant s under reasonable terms and conditions for the purpose of implementing the standard."
	Early disclosure to the Working Group of patent information that might be relevant to the standard is essential to re duce the possibility for delays in the development process and increase the likelihood that the draft publication will be approved for publication. Please notify the Chair < <u>mailto:r.b.marks@ieee.org</u> > as early as possible, in written or electronic form, of any patents (granted or under application) that may cover technology that is under consideration by or has been approved by IEEE 802.16. The Chair will disclose this notification via the IEEE 802.16 web site < <u>h</u> ttp://ieee802.org/16/ipr/patents/notices>.

#### 8.4.9.2.4 Low Density Parity Check Code (optional)

#### 8.4.9.2.4.1 Code Description

The fundamental LDPC code is a systematic linear block code with  $(N_c, N_k)$ , rate =  $N_k / N_c$  where  $N_c$  is length of code and  $N_k$  is information bit size. There is some code definition for the system and variety code rates may be constructed for good code performance for each code rate. Changing the consistent matrix size accommodates varying data field lengths. Explanation about consistent matrix is in Packet Encoding section.

#### 8.4.9.2.4.2 LDPC encoding

In a general analysis, an  $(N_c, N_k)$  LDPC code has  $N_k$  information bits and  $N_c$  coded bits with code rate  $r = N_k / N_c$ . . The parity-check matrix **H** is of dimension  $(N_c -$ 

 $N_k$ )× $N_c$ , and it defines a set of equations. For simplicity, let us put  $N_c$  -

 $N_k = N_p$ , where  $N_p$  denotes the number of parity bits.

We can get relation parity check matrix H between codewords v as belows

$$\boldsymbol{H}^{\cdot} \boldsymbol{v}^{t} = \boldsymbol{\theta} \tag{1}$$

for all codewords v.

An example parity-check matrix is shown below for an LDPC code (8, 4) as well as the expanded parity-check equations

$$\boldsymbol{H} = \begin{bmatrix} 1 & 0 & 1 & 0 & 1 & 0 & 1 & 0 \\ 1 & 0 & 0 & 1 & 0 & 1 & 0 & 1 \\ 0 & 1 & 1 & 0 & 0 & 1 & 1 & 0 \\ 0 & 1 & 0 & 1 & 1 & 0 & 0 & 1 \end{bmatrix} \Leftrightarrow \begin{cases} v_1 + v_3 + v_5 + v_7 = 0 \\ v_1 + v_4 + v_6 + v_8 = 0 \\ v_2 + v_3 + v_6 + v_7 = 0 \\ v_2 + v_4 + v_5 + v_8 = 0 \end{cases}$$

For efficient encoding of LDPC, *H* are divided into the form

$$\boldsymbol{H} = \begin{pmatrix} \boldsymbol{A} & \boldsymbol{B} & \boldsymbol{T} \\ \boldsymbol{C} & \boldsymbol{D} & \boldsymbol{E} \end{pmatrix}$$
(2)

where A is  $(N_p - g) \times N_k$ , B is  $(N_p - g) \times g$ , T is  $(N_p - g) \times (N_p - g)$ , C is  $g \times N_k$ , D is  $g \times g$ , and finally , E is  $g \times (N_p - g)$ . Further, all these matrices are sparse and T is lower triangular with ones along the diagon al.

Let  $v=(u, p_1, p_2)$  that u denotes the systematic part,  $p_1$  and  $p_2$  combined denote the parity part,  $p_1$  has length g, a nd  $p_2$  has length  $(N_p - g)$ . The definition equation  $H \cdot v^t = 0$  splits into two equations, as in equation 3 and 4 namely

$$\boldsymbol{A}\boldsymbol{u}^{T} + \boldsymbol{B}\boldsymbol{p}_{1}^{T} + \boldsymbol{T}\boldsymbol{p}_{2}^{T} = \boldsymbol{0}$$
 (3)

and

$$\left(-\boldsymbol{E}\boldsymbol{T}^{-1}\boldsymbol{A}+\boldsymbol{C}\right)\boldsymbol{u}^{T}+\left(-\boldsymbol{E}\boldsymbol{T}^{-1}\boldsymbol{B}+\boldsymbol{D}\right)\boldsymbol{p}_{1}^{T}=\boldsymbol{0}$$
(4)

Define  $\phi := -ET^{-1}B + D$  and when we use the parity check matrix as indicated appendix we can get  $\phi = I$ . The n from (4) we conclude that

$$\boldsymbol{p}_{1}^{T} = \left(-\boldsymbol{E}\boldsymbol{T}^{-1}\boldsymbol{A} + \boldsymbol{C}\right)\boldsymbol{u}^{T}$$
(5)

and

$$\boldsymbol{p}_{2}^{T} = \boldsymbol{T}^{-1} \left( \boldsymbol{A} \boldsymbol{u}^{T} + \boldsymbol{B} \boldsymbol{p}_{1}^{T} \right).$$
 (6)

As a result, the encoding procedures and the corresponding operations can be summarized below and illustrated in Fig. 1.

Encoding procedure

- Step 1) Compute  $Au^{T}$  and  $Cu^{T}$ .
- Step 2) Compute  $\boldsymbol{ET}^{-1}(\boldsymbol{Au}^{T})$ .
- Step 3) Compute  $\boldsymbol{p}_1^T$  by  $\boldsymbol{p}_1^T = \boldsymbol{E}\boldsymbol{T}^{-1}(\boldsymbol{A}\boldsymbol{u}^T) + \boldsymbol{C}\boldsymbol{u}^T$ .
- Step 4) Compute  $p_2^T$  by  $Tp_2^T = Au^T + Bp_1^T$ .



Fig. 1 Block diagram of the encoder architecture for the block LDPC code.

A detailed description of the *A*, *B*, *T*, *C*, *D* and *E* matrices of the code is contained in the Appendix LDPC Cod e Definition. The code is fully described by the definition of the H matrix as indicated in the appendix.

### 8.4.9.2.4.3 Rate Adjustment

As the code is rate flexible we can easily construct H matrix for various code rates (e.g. 1/2, 2/3, 3/4 and 5/6). F or each code rate, we design fundamental LDPC codes that achieve good performance whose H matrix describe d in appendix.

#### 8.4.9.2.4.4 Packet Encoding

In this section, we describe construction of parity check matrix of block LDPC codes to obtain packet data leng th flexibility. A block LDPC code is an almost structured LDPC codes whose parity-check matrix consists of small square blocks which are the zero matrix or a circulant permutation matrix. Let P be the  $N_s \times N_s$  permutati on matrix given by

$$\boldsymbol{P} = \begin{bmatrix} 0 & 1 & 0 & & 0 \\ 0 & 0 & 1 & \cdots & 0 \\ \vdots & \vdots & \vdots & & \vdots \\ 0 & 0 & 0 & \cdots & 1 \\ 1 & 0 & 0 & 0 & 0 \end{bmatrix}$$

Note that  $\mathbf{P}^i$  is just the circulant matrix of the identity matrix  $\mathbf{I}$  to the right by  $(i \mod N_s)$  times for any integer *i*. For simple notation,  $\mathbf{P}^{\infty}$  denotes the zero matrix.

Let **H** be the  $mN_s \times nN_s$  matrix defined by

$$H = \begin{vmatrix} P^{a_{11}} & P^{a_{12}} & P^{a_{13}} & \cdots & P^{a_{1(n-1)}} & P^{a_{1n}} \\ P^{a_{21}} & P^{a_{22}} & P^{a_{23}} & \cdots & P^{a_{2(n-1)}} & P^{a_{2n}} \\ P^{a_{31}} & P^{a_{32}} & P^{a_{33}} & \cdots & P^{a_{3(n-1)}} & P^{a_{3n}} \\ \vdots & \vdots & \vdots & \cdots & \vdots & \vdots \\ P^{a_{m1}} & P^{a_{m2}} & P^{a_{m3}} & \cdots & P^{a_{m(n-1)}} & P^{a_{mn}} \end{vmatrix}$$

where  $a_{ij} \in \{0, 1, \dots, N_s - 1, \infty\}$ . When *H* has full rank, then its codeword size  $N_c$  is  $nN_s$  and information bit size  $N_k$  is  $(n-m)N_s$ . Therefore, its code rate is given by

$$R = \frac{N_s n - N_s m}{N_s n} = \frac{n - m}{n} = 1 - \frac{m}{n}$$

regardless of its block length  $nN_s$ .

Therefore, we can obtain larger size block LDPC codes by increasing the size of circulant permutation matrices P which is an element matrix of H matrix. Also, we can straightforwardly get small size block LDPC codes by decreasing the size of P.

For example, when we use  $N_s = 1$  with code rate 1/2 case in appendix, we can construct (24, 12) block LDPC c odes and for the same case. If we set  $N_s = 10$ , then we can obtain (240, 120) block LDPC codes.

### Appendix (normative); LDPC Code Definition

A full definition of an LDPC code can be accomplished through identification of the locations of the "edges" b etween the variable nodes (codeword bits) and check nodes (parity relationships). Figure 2 shows a Tanner gra ph of an example LDPC code, depicting the arrangement of the check nodes, variable nodes, and the "edges" c onnecting them.



Fig. 2 This is an example Tanner graph of an LDPC code, showing the check nodes, variable nodes, and edges. The codeword is made up of the bits represented by the variable nodes. In this case the codeword has eight bits.

Each check node represents a parity relationship between the codeword bits represented by the variable nodes c onnected to it by the edges. The number of edges connected to a check node is the "degree" of the check node, and the number of edges connected to a variable node is the "degree" of the variable node. For the specified co de all check nodes are of degree eighteen, all variable nodes related to the systematic information bits are of de gree four, and all variable nodes corresponding to parity bits are of degree two except for the last, which is of d egree one.

This LDPC code list file contains six parts to describe the parity check matrix H. When matrix H split into the f orm

$$H = \begin{pmatrix} A & B & T \\ C & D & E \end{pmatrix}$$

we describe the matrices A, B, T, C, D and E.

An example for a  $(8 \times N_s, 4 \times N_s)$  code with n=8, m=4, we design the parity check matrix

$$H = \begin{bmatrix} P^{0} & 0 & 0 & P^{0} & P^{1} & P^{0} & 0 & 0 \\ 0 & P^{1} & P^{4} & 0 & 0 & P^{0} & P^{0} & 0 \\ 0 & P^{2} & 0 & P^{6} & P^{2} & 0 & P^{0} & P^{0} \\ P^{3} & 0 & P^{5} & 0 & P^{3} & 0 & 0 & P^{0} \end{bmatrix}$$

where  $\boldsymbol{\theta}$  is  $N_s \times N_s$  zero matrix.

Then we describe matrices A, B, T, C, D and E as below

$$\boldsymbol{A} = \begin{bmatrix} 0 & \infty & \infty & 0 \\ \infty & 1 & 4 & \infty \\ \infty & 2 & \infty & 6 \end{bmatrix}, \quad \boldsymbol{B} = \begin{bmatrix} 1 \\ \infty \\ 2 \end{bmatrix}, \quad \boldsymbol{T} = \begin{bmatrix} 0 & \infty & \infty \\ 0 & 0 & \infty \\ \infty & 0 & 0 \end{bmatrix}$$
$$\boldsymbol{C} = \begin{bmatrix} 3 & \infty & 5 & \infty \end{bmatrix}, \quad \boldsymbol{D} = \begin{bmatrix} 3 \end{bmatrix}, \quad \boldsymbol{E} = \begin{bmatrix} \infty & \infty & 0 \end{bmatrix}$$

### Code rate = 1/2



# $\mathbf{E} = \begin{bmatrix} \infty & 1 \end{bmatrix}$

Code Rate = 2/30 0 0 0 0 0 ∞ ∞ ∞ ∞ ∞ ∞ ∞ ∞ ∞ ∞  $\infty \infty \infty \infty$  $\infty$ s  $\infty$  $0\quad 0\quad 0\quad \infty\quad \infty\quad \infty\quad \infty\quad \infty\quad \infty\quad \infty\quad \infty\quad \infty\quad 0\quad 0\quad 0\quad \infty\quad \infty$  $\infty$ 00 x x **0 0 0 0** ∞ ∞ ∞ ∞ **7** ∞ ∞ ∞ ∞ ∞  $\infty$  $\infty \infty \infty \infty \infty \infty$ 0 0 00 

 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 0
 0
 0
 0
 ∞
 3
 5
 ∞
 1
 ∞
 ∞
 15

 ∞
 ∞
 ∞
 1
 ∞
 ∞
 4
 ∞
 ∞
 ∞
 ∞
 ∞
 3
 5
 ∞
 1
 ∞
 ∞
 15

 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞
 ∞</ 6 ∞ ∞ ∞ ∞ ∞ ∞ ∞ ∞ 6 ∞ ∞ 2 ∞ 15 ∞ ∞ ∞ ∞ 14 8  $\infty$ A =  $\infty$ 3 ∞ ∞ ∞ ∞ ∞ ∞ ∞ 6 ∞ 7 ∞ 5 ∞ ∞ 9 ∞ 11 ∞ ∞ 3 ∞ 15 ∞ x ∞ 6 ∞ 1 ∞ 2 ∞ ∞ ∞ ∞ ∞ ∞ ∞ ∞ ∞ ∞ ∞ 8 ∞ ∞ 12 ∞ 9 4 x  $\infty \quad \mathbf{3} \quad \boldsymbol{\infty} \quad \boldsymbol{\infty} \quad \boldsymbol{\infty} \quad \boldsymbol{\infty} \quad \boldsymbol{\infty} \quad \mathbf{12} \quad \boldsymbol{\infty} \quad \boldsymbol{\infty} \quad \boldsymbol{\infty} \quad \mathbf{3} \quad \boldsymbol{\infty} \quad \mathbf{14} \quad \boldsymbol{\infty} \quad \mathbf{2} \quad \mathbf{14} \quad \boldsymbol{\infty} \quad \mathbf{11} \quad \boldsymbol{\infty}$ 1  $\infty$ 6 ∞ ∞ ∞ ∞ 8 ∞ 15 ∞ ∞ ∞ ∞ ∞ 9 ∞ ∞ 4 13 s 13  $\infty \infty \infty \infty$  $\infty$  $\infty$  $\infty$  $\infty$ 7  $\infty$  10  $\infty$   $\infty$  5  $\infty$   $\infty$  6  $\infty$  $\infty$ **13** ∞  $\infty$ 14 9 0  $\infty$  $\infty$  $\infty$  $\infty$ в =  $\infty$ Ns-1  $\infty$  $\infty$  $\infty$  $\infty$ 0  $\infty$  $\infty$  $\infty$ s  $\infty$ s s  $\infty$  $\infty$ 0 0  $\infty$  $\infty$  $\infty$  $\infty$  $\infty \infty$  $\infty$  $\infty$  $\infty$ s 0 0  $\infty \infty \infty$  $\infty$  $\infty$ x 00  $\infty$  $\infty$ x s  $\infty$  $\infty \quad \infty \quad \infty \quad \infty \quad 0 \quad 0 \quad \infty \quad \infty \quad \infty$ x т =  $\infty$  $\infty \quad \infty \quad \infty \quad \mathbf{0} \quad \mathbf{0} \quad \infty \quad \infty \quad \infty \quad \infty$  $\infty$  $\infty$  $\infty \quad \infty \quad \infty \quad \mathbf{0} \quad \mathbf{0} \quad \infty \quad \infty \quad \infty$ s  $\infty$  $\infty \quad \infty \quad 0 \quad 0 \quad \infty \quad \infty \quad \infty \quad \infty \quad \infty \quad \infty$ <sup>o</sup>  $\infty$  $\infty$  $\infty \quad \textbf{0} \quad \textbf{0} \quad \infty \quad \infty \quad \infty \quad \infty \quad \infty$  $\infty$  $\infty$ 0  $\infty$  $\infty$  $\mathbf{0} \quad \infty \quad \infty \quad \infty \quad \infty \quad \mathbf{0}$  $\infty$  $\infty$ 0 0 x ø  $\infty \infty \infty \infty$  $\infty$ x  $\infty$ C = ∞ 4 ∞ 2 ∞ ∞ ∞ 5 ∞ ∞ ∞ ∞ 7 ∞ ∞ ∞ 3 ∞ 13 15 ∞ ∞ 10 ∞ D = 1

 $\mathbf{E} = \begin{bmatrix} \infty & \mathbf{1} \end{bmatrix}$ 

Code	Code Rate = 3/4																							
	ſ	0	0	0	0	0	0	$\infty$	0	0	0	0	$\infty$	$\infty$										
		$\infty$	$\infty$	$\infty$	$\infty$	$\infty$	$\infty$	0	0	0	0	0	0	$\infty$	$\infty$	$\infty$	$\infty$	$\infty$	2	x	x	s	0	0
		$\infty$	5	$\infty$	9	$\infty$	0	0	0	0	0	$\infty$	1	$\infty$	$\infty$	8	$\infty$							
A =	=	2	$\infty$	$\infty$	$\infty$	3	$\infty$	$\infty$	13	$\infty$	$\infty$	15	$\infty$	2	$\infty$	$\infty$	10	$\infty$	0	$\infty$	7	5	$\infty$	4
		$\infty$	$\infty$	$\infty$	$\infty$	$\infty$	4	$\infty$	$\infty$	12	11	$\infty$	10	7	$\infty$	17	$\infty$	$\infty$	$\infty$	13	$\infty$	15	14	$\infty$
		$\infty$	$\infty$	1	6	$\infty$	$\infty$	3	4	$\infty$	$\infty$	2	$\infty$	$\infty$	$\infty$	13	$\infty$	11	2	$\infty$	5	$\infty$	7	$\infty$
		$\infty$	8	$\infty$	$\infty$	10	3	$\infty$	$\infty$	$\infty$	12	$\infty$	7	$\infty$	5	$\infty$	$\infty$	$\infty$	$\infty$	$\infty$	$\infty$	13	15	14
			_																					
		8																						
		0																						
в =		10																						
	=	8																						
		8																						
		8																						
		11																						
	_	-	_																					
	ľ	0	0	x	$\infty$	$\infty$	x	$\infty$	œ															
		x	0	0	x	$\infty$	x	s	œ	,														
		8	$\infty$	0	0	$\infty$	x	s	x	,														
т :	=	8	$\infty$	x	0	0	x	s	x	,														
		Ns-1	$\infty$	x	x	0	0	s	œ	,														
		8	$\infty$	00	$\infty$	$\infty$	0	0	œ	,														
		8	$\infty$	00	$\infty$	$\infty$	x	0	0															
										•														
С =	= [	11	$\infty$	6	$\infty$	$\infty$	$\infty$	9	$\infty$	16	$\infty$	$\infty$	$\infty$	$\infty$	1	$\infty$	5	6	$\infty$	14	7	$\infty$	x	$\infty$
D :	= [	2	I																					
E :	- [	0	x	x	x	×	x	x	0	Ι														

# 8

#### Code rate = 5/6



<u>Supplementary information on the proposed text:</u> Simulation results of Intel LDPC coding vs. Samsung B-LDPC coding.

The following two graphs in Figure 1 and 2 highlights the performance comparison of the two proposed LDPC encoders, i.e., Intel LDPC (illustrated in IEEE 802.11-03/0865r1) and Samsung B-LDPC (proposed in this contribution).



Fig. 1. Performance comparison between a (792,396) B-LDPC code and a (800,400) Intel LDPC code

The performance of a BLDPC code with column is compared with that of a Intel LDPC code frame leng th 800 in Fig. 1, in terms of FER (frame error rate). The parity check matrix of BLDPC code has the length of th e B-LDPC code is 792 and the code rate 1/2.



Fig. 2 Performance comparison between a (1200,800) B-LDPC code and a (1200,800) Intel LDPC code

The performance of a BLDPC code with column is compared with that of a Intel LDPC code frame length 12 00 in Fig. 2, in terms of FER (frame error rate). The parity check matrix of BLDPC code has the length of the B -LDPC code is 1200 and the code rate 2/3.